



SGM8541/2/4

1.1MHz, 42µA, Rail-to-Rail I/O

CMOS Operational Amplifiers

PRODUCT DESCRIPTION

The SGM8541(single), SGM8542(dual) and SGM8544(quad) are rail-to-rail input and output voltage feedback amplifiers offering low cost. They have a wide input common-mode voltage range and output voltage swing, and take the minimum operating supply voltage down to 2.1V and the maximum recommended supply voltage is 5.5V. All are specified over the extended -40°C to +125°C temperature range.

The SGM8541/2/4 provide 1.1MHz bandwidth at a low current consumption of 42µA per amplifier. Very low input bias currents of 0.5pA enable SGM8541/2/4 to be used for integrators, photodiode amplifiers, and piezoelectric sensors. Rail-to-rail inputs and outputs are useful to designers buffering ASIC in single-supply systems.

Applications for these amplifiers include safety monitoring, portable equipment, battery and power supply control, and signal conditioning and interfacing for transducers in very low power systems.

The SGM8541 is available in the tiny SOT23-5, SC70-5 and SO-8 packages. The SGM8542 comes in the miniature SO-8 and MSOP-8 packages. The SGM8544 is offered in TSSOP-14 and SO-14 packages.

APPLICATIONS

ASIC Input or Output Amplifier

Sensor Interface

Piezo Electric Transducer Amplifier

Medical Instrumentation

Mobile Communication

Audio Output

Portable Systems

Smoke Detectors

Mobile Telephone

Notebook PC

PCMCIA Cards

Battery –Powered Equipment

FEATURES

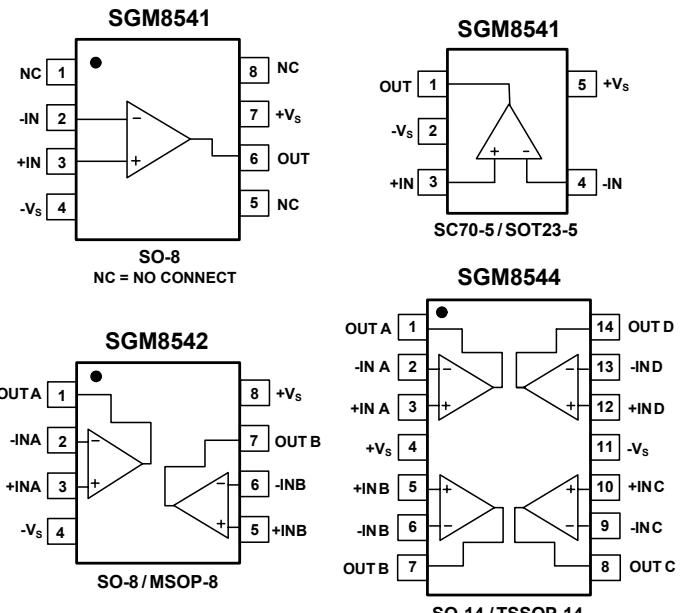
- Low Cost
- Rail-to-Rail Input and Output
0.8mV Typical V_{os}
- Unity Gain Stable
- Gain Bandwidth Product: 1.1MHz
- Very Low Input Bias Current: 0.5pA
- Operates on 2.1V to 5.5V Supplies
- Input Voltage Range = -0.1V to +5.6V with $V_s = 5.5V$
- Low Supply Current: 42µA/Amplifier
- Small Packaging

SGM8541 Available in SO-8, SC70-5 and SOT23-5

SGM8542 Available in SO-8 and MSOP-8

SGM8544 Available in SO-14 and TSSOP-14

PIN CONFIGURATIONS (Top View)



PACKAGE/ORDERING INFORMATION

MODEL	ORDER NUMBER	PACKAGE DESCRIPTION	PACKAGE OPTION	MARKING INFORMATION
SGM8541	SGM8541XC5/TR	SC70-5	Tape and Reel, 3000	8541
	SGM8541XN5/TR	SOT23-5	Tape and Reel, 3000	8541
	SGM8541XS/TR	SO-8	Tape and Reel, 2500	SGM8541XS
SGM8542	SGM8542XS/TR	SO-8	Tape and Reel, 2500	SGM8542XS
	SGM8542XMS/TR	MSOP-8	Tape and Reel, 3000	SGM8542XMS
SGM8544	SGM8544XS14/TR	SO-14	Tape and Reel, 2500	SGM8544XS14
	SGM8544XTS14/TR	TSSOP-14	Tape and Reel, 3000	SGM8544XTS14

Note: SC70-5 package is same as SOT-353 package.

ABSOLUTE MAXIMUM RATINGS

Supply Voltage, V+ to V-	7.5V	SOT23-5, θ_{JA}	190°C/W
Common-Mode Input Voltage	(-V _S) - 0.5 V to (+V _S) +0.5V	SO-8, θ_{JA}	125°C/W
Storage Temperature Range	-65°C to +150°C	MSOP-8, θ_{JA}	216°C/W
Junction Temperature	150°C	Lead Temperature Range (Soldering 10 sec)	260°C
Operating Temperature Range	-55°C to +150°C	ESD Susceptibility	
Package Thermal Resistance @ T _A = 25°C		HBM	4000V
SC70-5, θ_{JA}	333°C/W	MM	400V

NOTES

1. Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only; functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

CAUTION

This integrated circuit can be damaged by ESD if you don't pay attention to ESD protection. SGMICRO recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

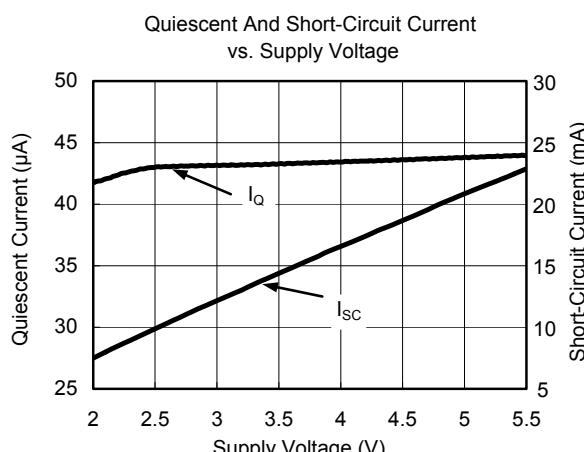
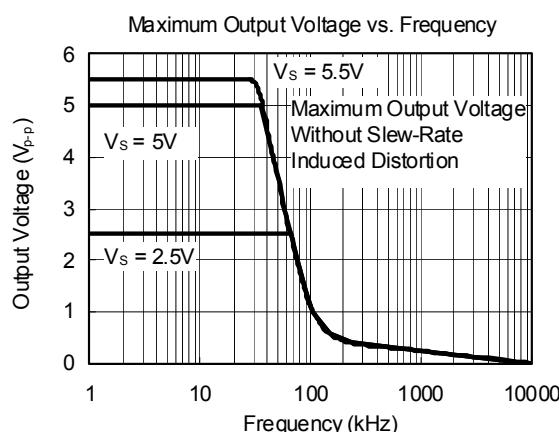
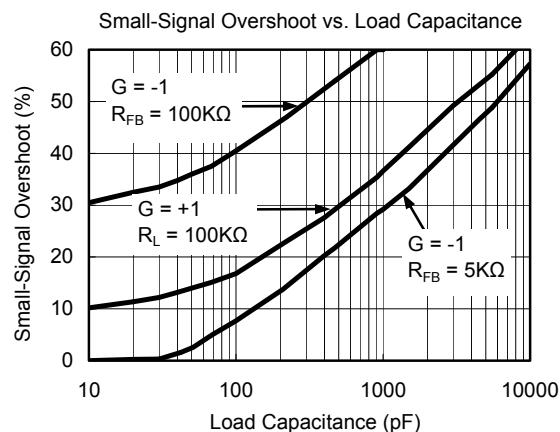
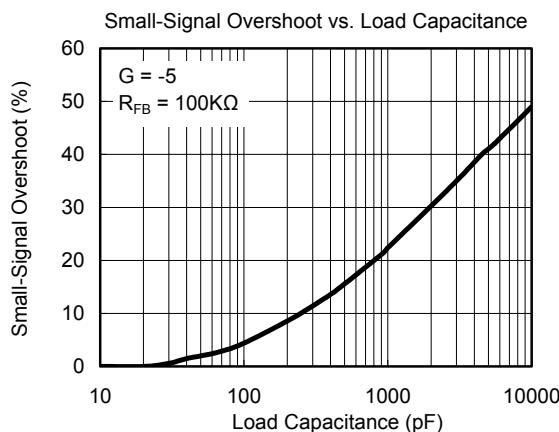
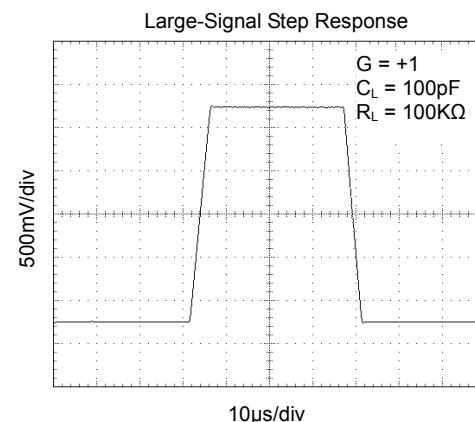
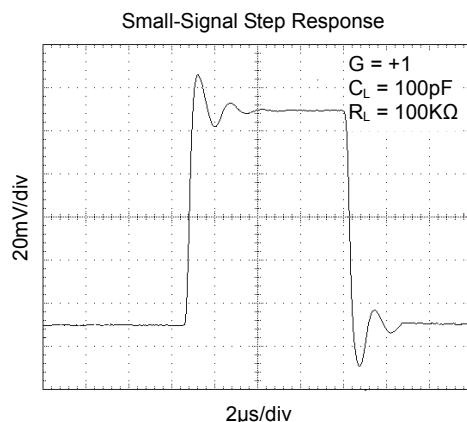
SGM8541/2/4

**1.1MHz, 42 μ A, Rail-to-Rail I/O
CMOS Operational Amplifiers**

ELECTRICAL CHARACTERISTICS: $V_S = +5V$ (At $R_L = 100k\Omega$ connected to $V_S/2$, and $V_{OUT} = V_S/2$, unless otherwise noted.)

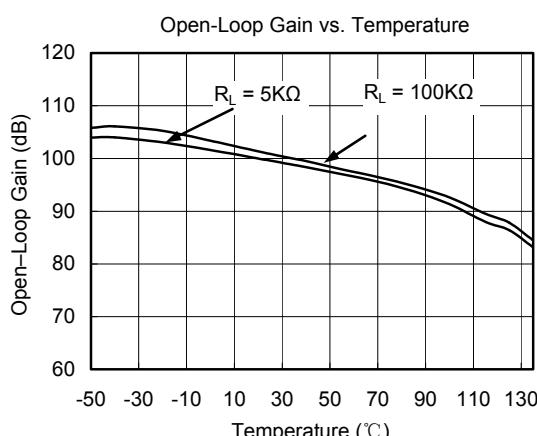
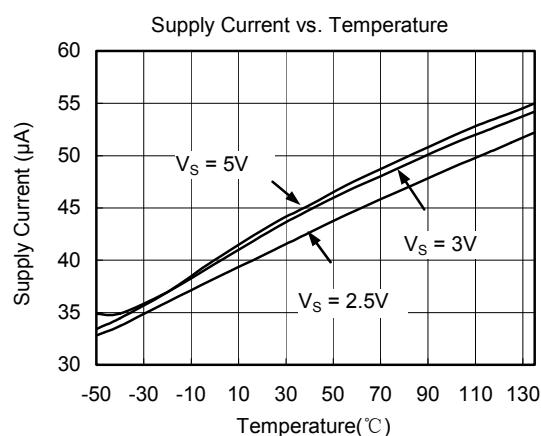
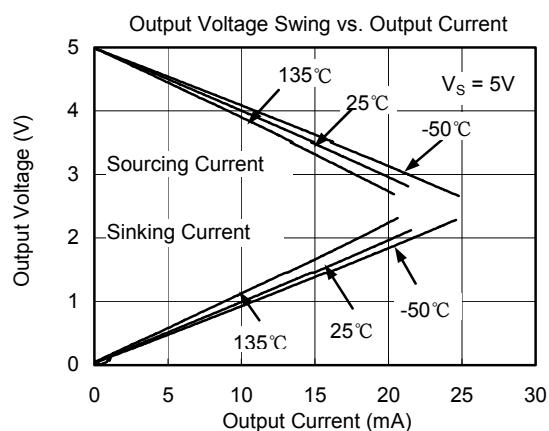
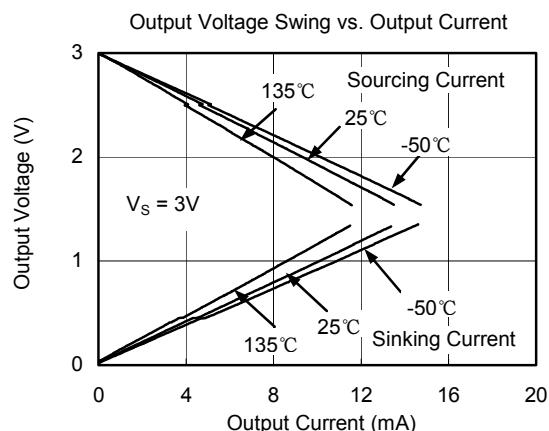
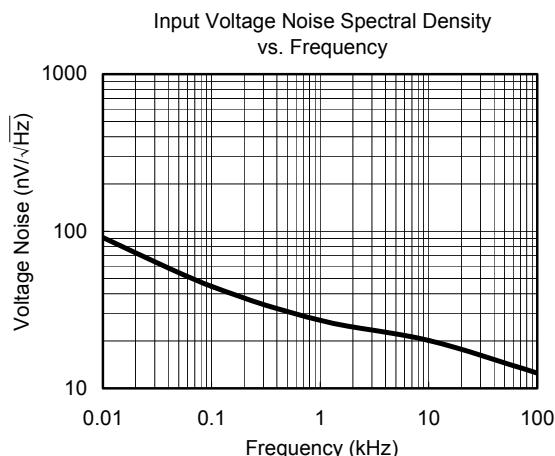
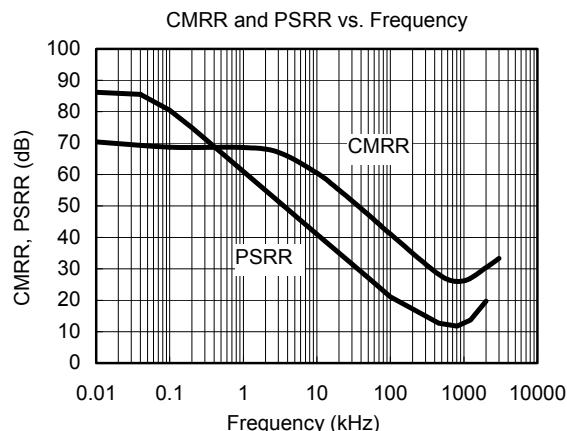
PARAMETER	CONDITIONS	SGM8541/2/4						
		TYP	MIN/MAX OVER TEMPERATURE					
		+25°C	+25°C	0°C to 70°C	-40°C to 85°C	-40°C to 125°C	UNITS	MIN/ MAX
INPUT CHARACTERISTICS								
Input Offset Voltage (V_{OS})		± 0.8	± 3.5	± 3.7	± 4.1	± 5.4	mV	MAX
Input Bias Current (I_B)		0.5					pA	TYP
Input Offset Current (I_{OS})		0.5					pA	TYP
Common-Mode Voltage Range (V_{CM})	$V_S = 5.5V$	-0.1 to +5.6					V	TYP
Common-Mode Rejection Ratio (CMRR)	$V_S = 5.5V, V_{CM} = -0.1V$ to 4V	88	72	72	72	72	dB	MIN
	$V_S = 5.5V, V_{CM} = -0.1V$ to 5.6V	78	60	59	59	57	dB	MIN
Open-Loop Voltage Gain (A_{OL})	$R_L = 5K\Omega, V_o = 0.1V$ to 4.9V	90	80	80	80	78	dB	MIN
	$R_L = 100K\Omega, V_o = 0.035V$ to 4.965V	94	85	85	85	82	dB	MIN
Input Offset Voltage Drift ($\Delta V_{OS}/\Delta T$)		2.7					$\mu V/^\circ C$	TYP
OUTPUT CHARACTERISTICS								
Output Voltage Swing from Rail							V	TYP
Output Current (I_{OUT})	$R_L = 100K\Omega$	0.008					mA	MIN
		23	20	19.1	18.8	18		
POWER SUPPLY								
Operating Voltage Range							V	MIN
			2.1	2.5	2.5	2.5	V	MAX
Power Supply Rejection Ratio (PSRR)	$V_s = +2.5V$ to + 5.5V		5.5	5.5	5.5	5.5	V	
	$V_{CM} = (-V_s) + 0.5$	92	76	74	71	70	dB	MIN
Quiescent Current / Amplifier (I_Q)	$I_{OUT} = 0$	42	54	59	60	64	μA	MAX
DYNAMIC PERFORMANCE								
Gain-Bandwidth Product (GBP)	$C_L = 100pF$						MHz	TYP
Slew Rate (SR)		1.1					V/ μ s	TYP
Settling Time to 0.1% (t_s)	G = +1, 2V Output Step	0.52					μ s	TYP
Overload Recovery Time	G = +1, 2V Output Step	5.3					μ s	TYP
	V_{IN} :Gain = V_s	2.6						
NOISE PERFORMANCE								
Voltage Noise Density (e_n)	$f = 1kHz$	27					nV/\sqrt{Hz}	TYP
	$f = 10kHz$	20					nV/\sqrt{Hz}	TYP

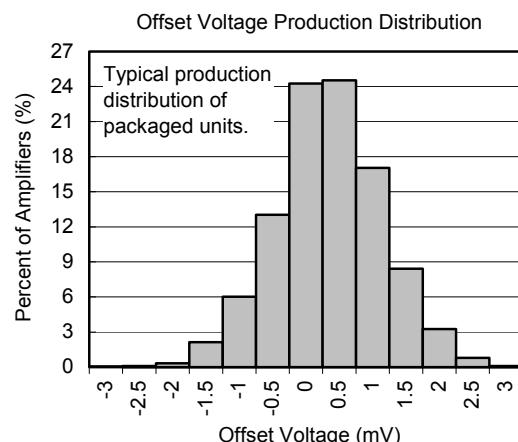
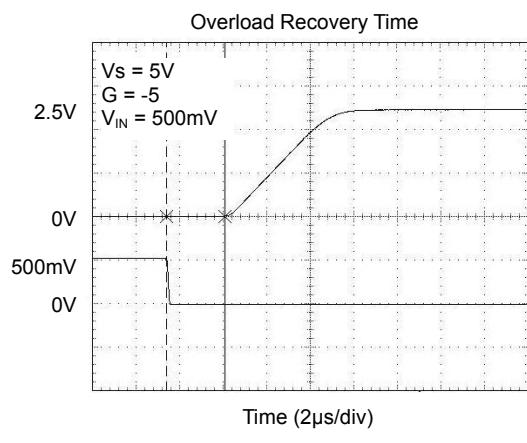
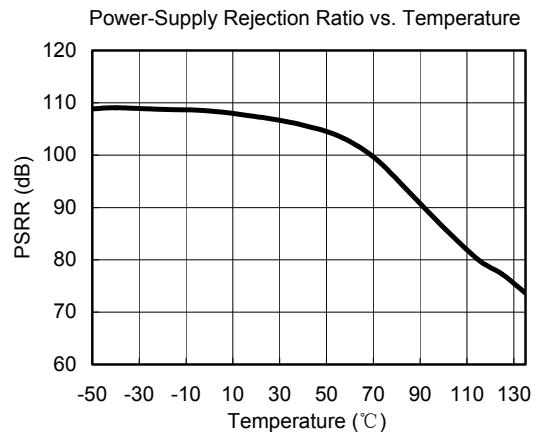
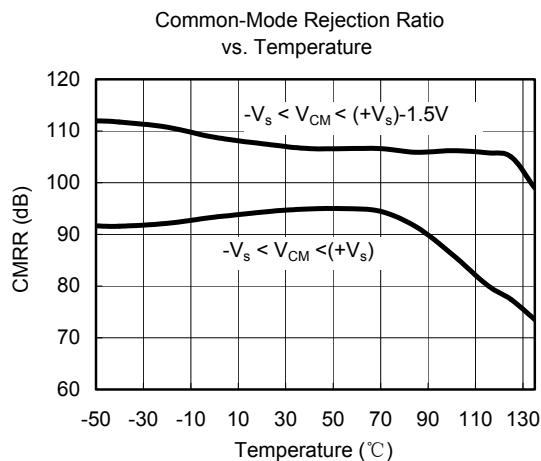
Specifications subject to change without notice.

TYPICAL PERFORMANCE CHARACTERISTICSAt $T_A = +25^\circ\text{C}$, $V_S = +5\text{V}$, and $R_L = 100\text{K}\Omega$ connected to $V_S/2$, unless otherwise noted.

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APPLICATION NOTES**Driving Capacitive Loads**

The SGM854X can directly drive 250pF in unity-gain without oscillation. The unity-gain follower (buffer) is the most sensitive configuration to capacitive loading. Direct capacitive loading reduces the phase margin of amplifiers and these results in ringing or even oscillation. Applications that require greater capacitive drive capability should use an isolation resistor between the output and the capacitive load like the circuit in Figure 1. The isolation resistor R_{ISO} and the load capacitor C_L form a zero to increase stability. The bigger the R_{ISO} resistor value, the more stable V_{OUT} will be. Note that this method results in a loss of gain accuracy because R_{ISO} forms a voltage divider with the R_{LOAD} .

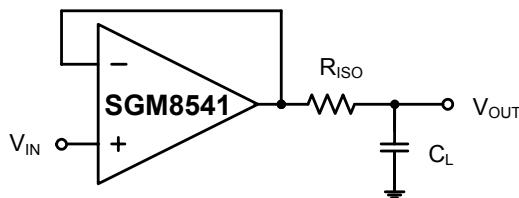


Figure 1. Indirectly Driving Heavy Capacitive Load

An improvement circuit is shown in Figure 2. It provides DC accuracy as well as AC stability. R_F provides the DC accuracy by connecting the inverting signal with the output, C_F and R_{ISO} serve to counteract the loss of phase margin by feeding the high frequency component of the output signal back to the amplifier's inverting input, thereby preserving phase margin in the overall feedback loop.

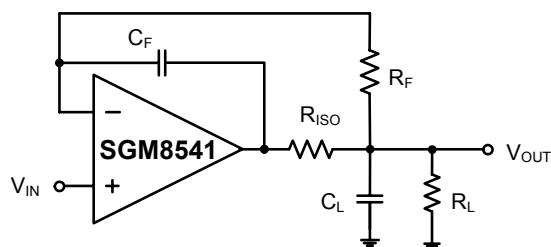


Figure 2. Indirectly Driving Heavy Capacitive Load with DC Accuracy

For no-buffer configuration, there are two others ways to increase the phase margin: (a) by increasing the amplifier's gain or (b) by placing a capacitor in parallel with the feedback resistor to counteract the parasitic capacitance associated with inverting node.

Power-Supply Bypassing and Layout

The SGM854X family operates from either a single +2.5V to +5.5V supply or dual $\pm 1.25V$ to $\pm 2.75V$ supplies. For single-supply operation, bypass the power supply V_{DD} with a 0.1 μ F ceramic capacitor which should be placed close to the V_{DD} pin. For dual-supply operation, both the V_{DD} and the V_{SS} supplies should be bypassed to ground with separate 0.1 μ F ceramic capacitors. 2.2 μ F tantalum capacitor can be added for better performance.

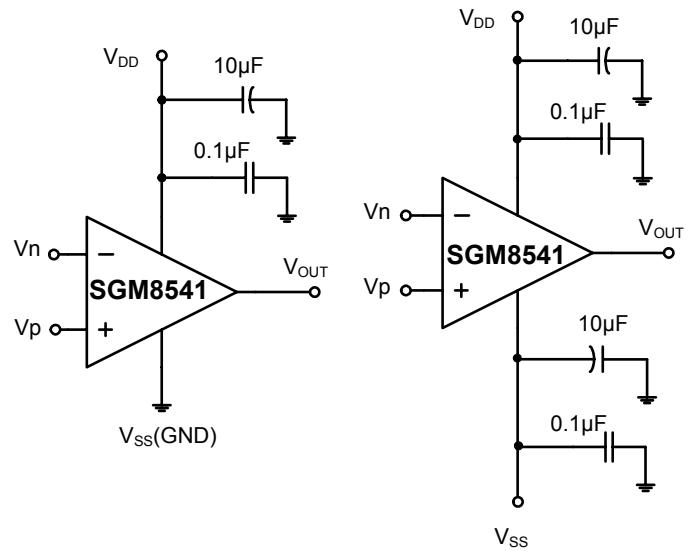


Figure 3. Amplifier with Bypass Capacitors

TYPICAL APPLICATION CIRCUITS**Differential Amplifier**

The circuit shown in Figure 4 performs the difference function. If the resistors ratios are equal ($R_4 / R_3 = R_2 / R_1$), then

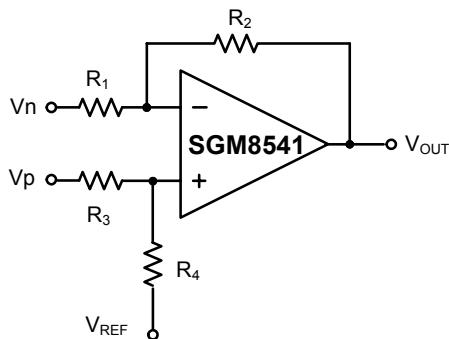
$$V_{OUT} = (V_p - V_n) \times R_2 / R_1 + V_{REF}$$


Figure 4. Differential Amplifier

Instrumentation Amplifier

The circuit in Figure 5 performs the same function as that in Figure 4 but with the high input impedance.

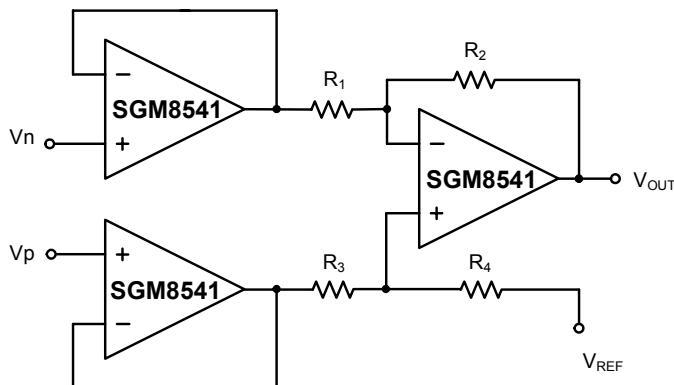


Figure 5. Instrumentation Amplifier

Low Pass Active Filter

The low pass filter shown in Figure 6 has a DC gain of $(-R_2 / R_1)$ and the $-3dB$ corner frequency is $1/2\pi R_2 C$. Make sure the filter is within the bandwidth of the amplifier. The Large values of feedback resistors can couple with parasitic capacitance and cause undesired effects such as ringing or oscillation in high-speed amplifiers. Keep resistors value as low as possible and consistent with output loading consideration.

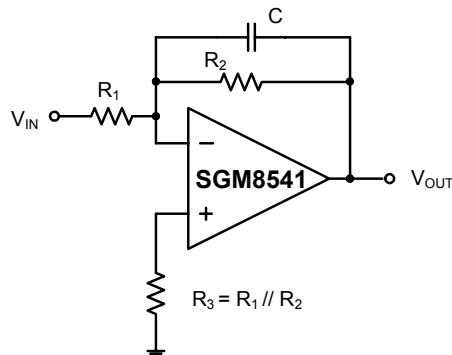
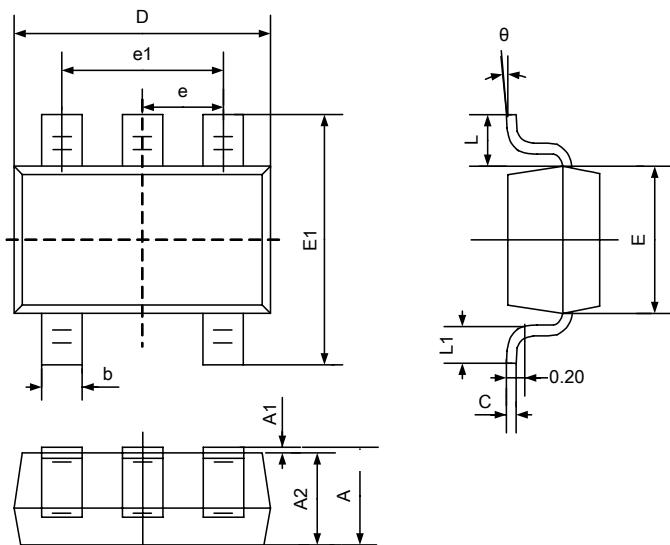


Figure 6. Low Pass Active Filter

PACKAGE OUTLINE DIMENSIONS

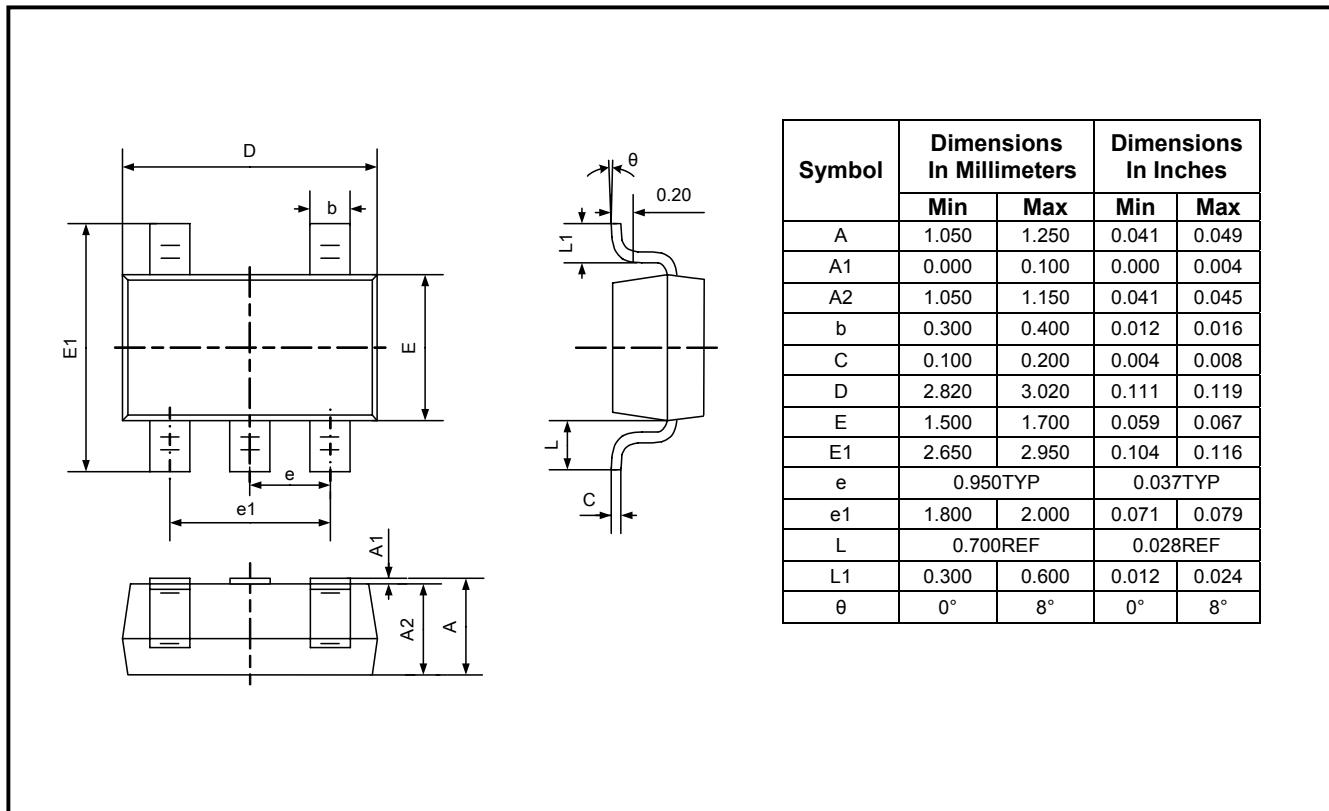
SC70-5 /SOT-353



Symbol	Dimensions In Millimeters		Dimensions In Inches	
	Min	Max	Min	Max
A	0.900	1.100	0.035	0.043
A1	0.000	0.100	0.000	0.004
A2	0.900	1.000	0.035	0.039
b	0.150	0.350	0.006	0.014
C	0.080	0.150	0.003	0.006
D	2.000	2.200	0.079	0.087
E	1.150	1.350	0.045	0.053
E1	2.150	2.450	0.085	0.096
e	0.650TYP		0.026TYP	
e1	1.200	1.400	0.047	0.055
L	0.525REF		0.021REF	
L1	0.260	0.460	0.010	0.018
θ	0°	8°	0°	8°

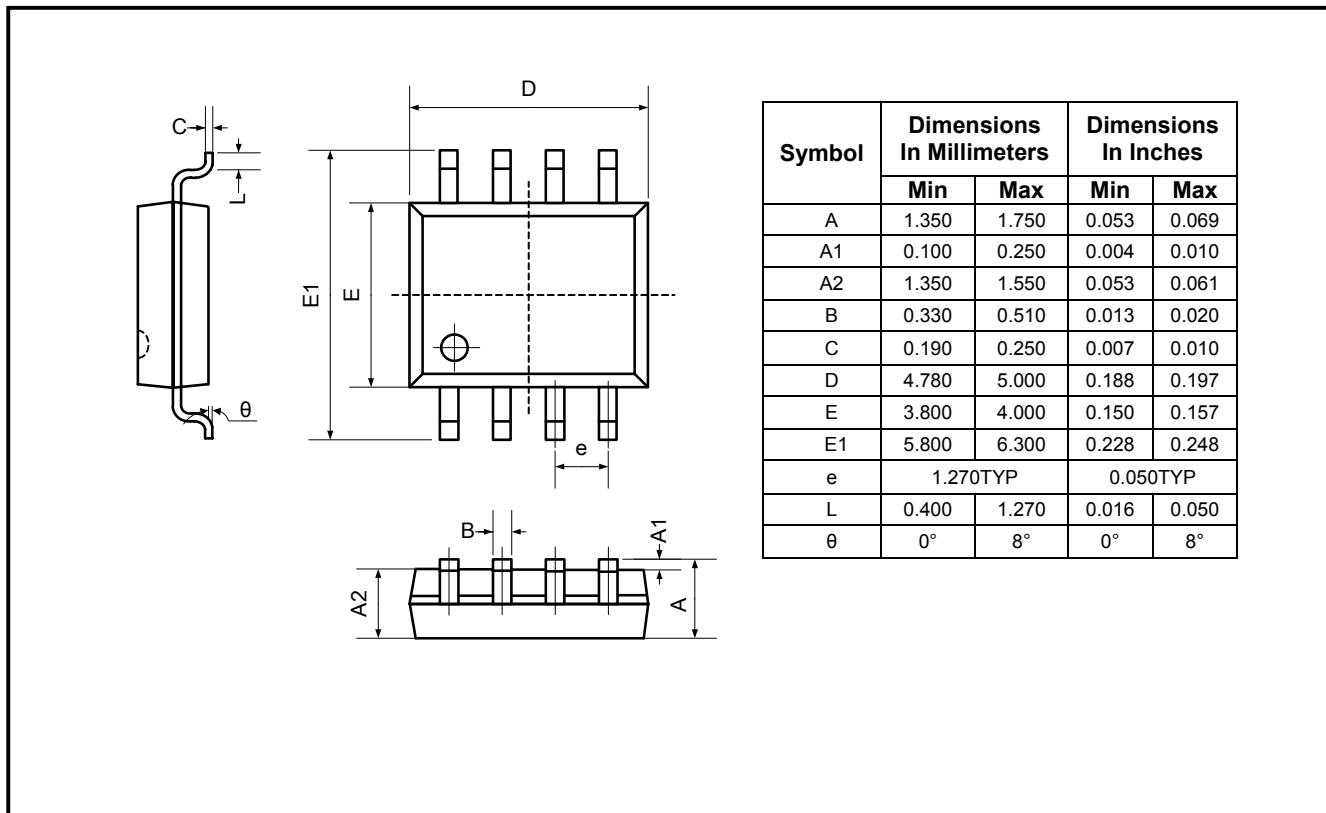
PACKAGE OUTLINE DIMENSIONS

SOT23-5



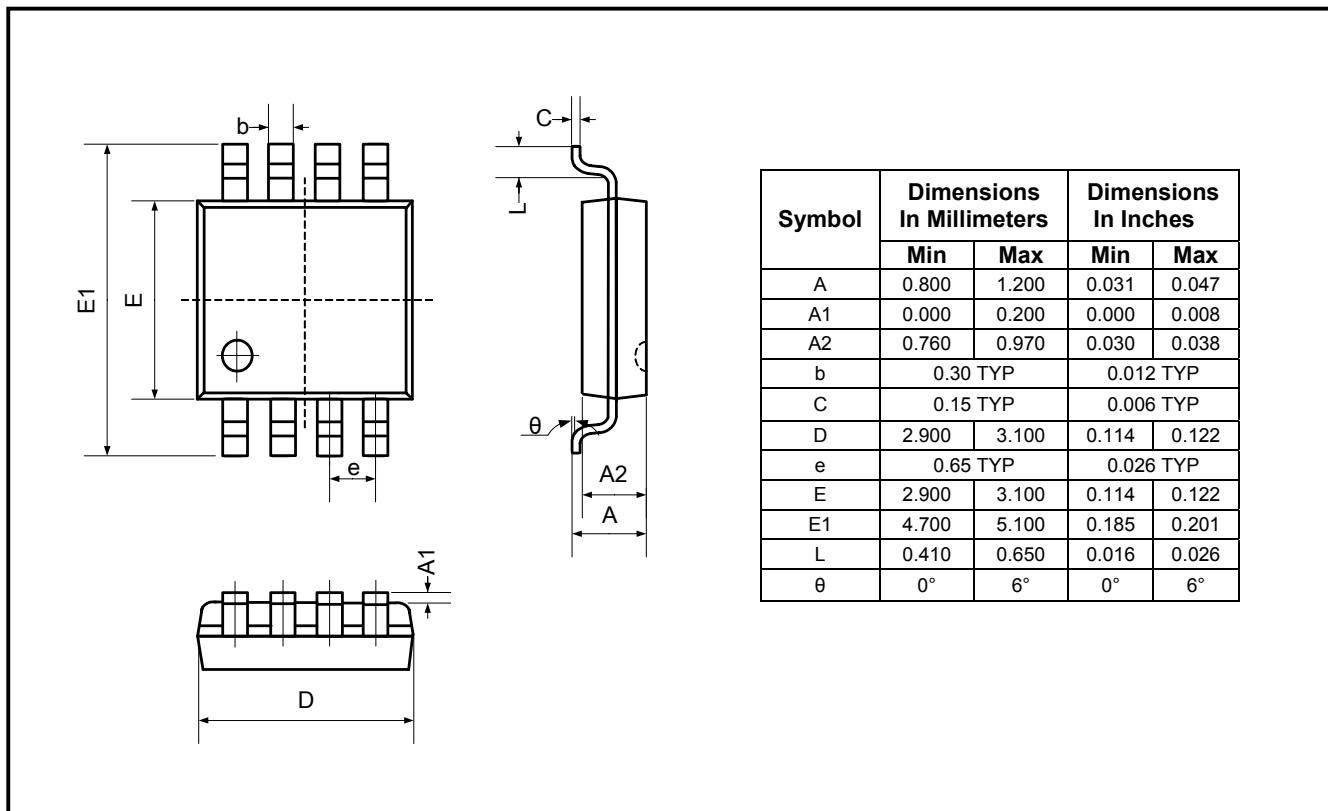
PACKAGE OUTLINE DIMENSIONS

SO-8



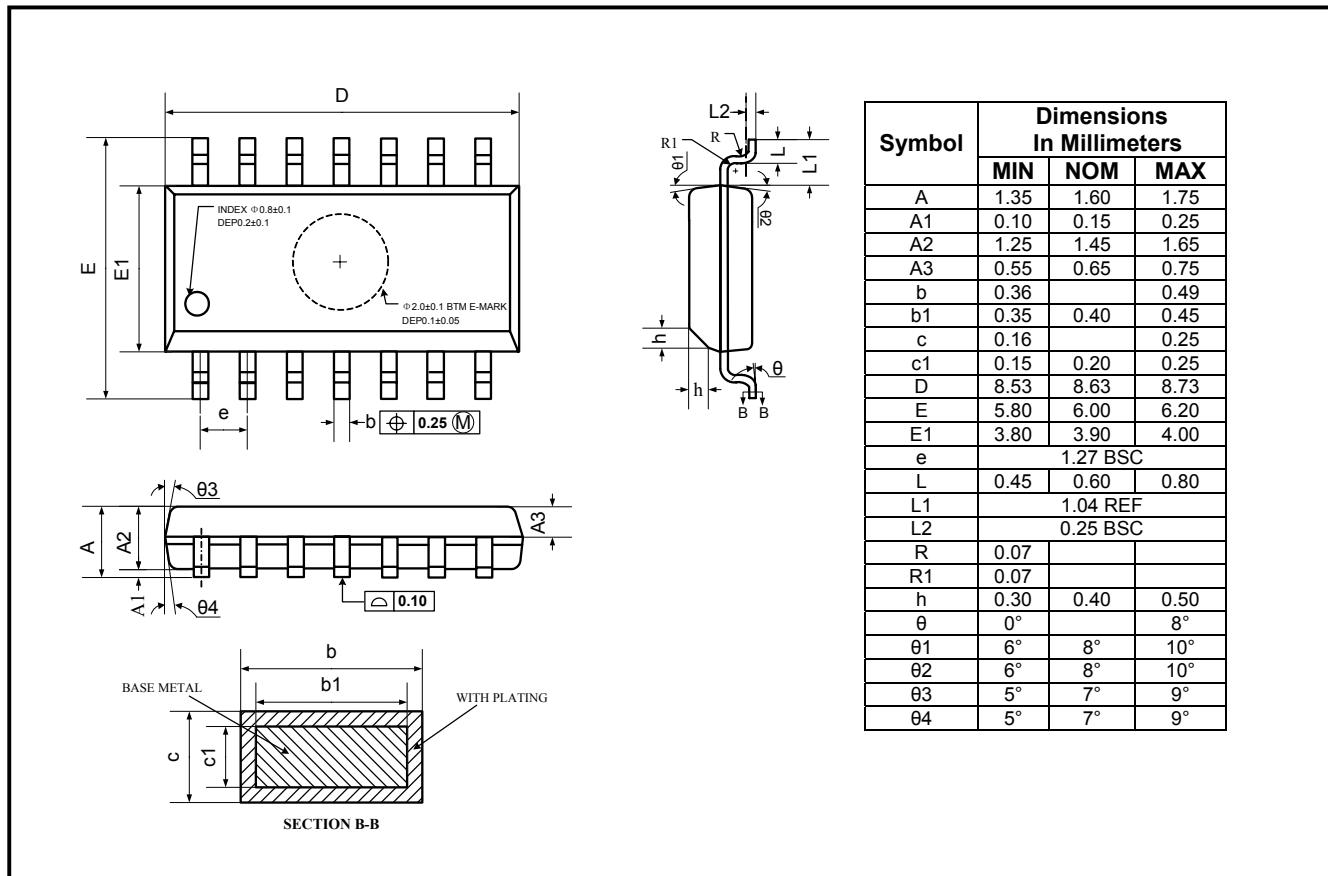
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MSOP-8



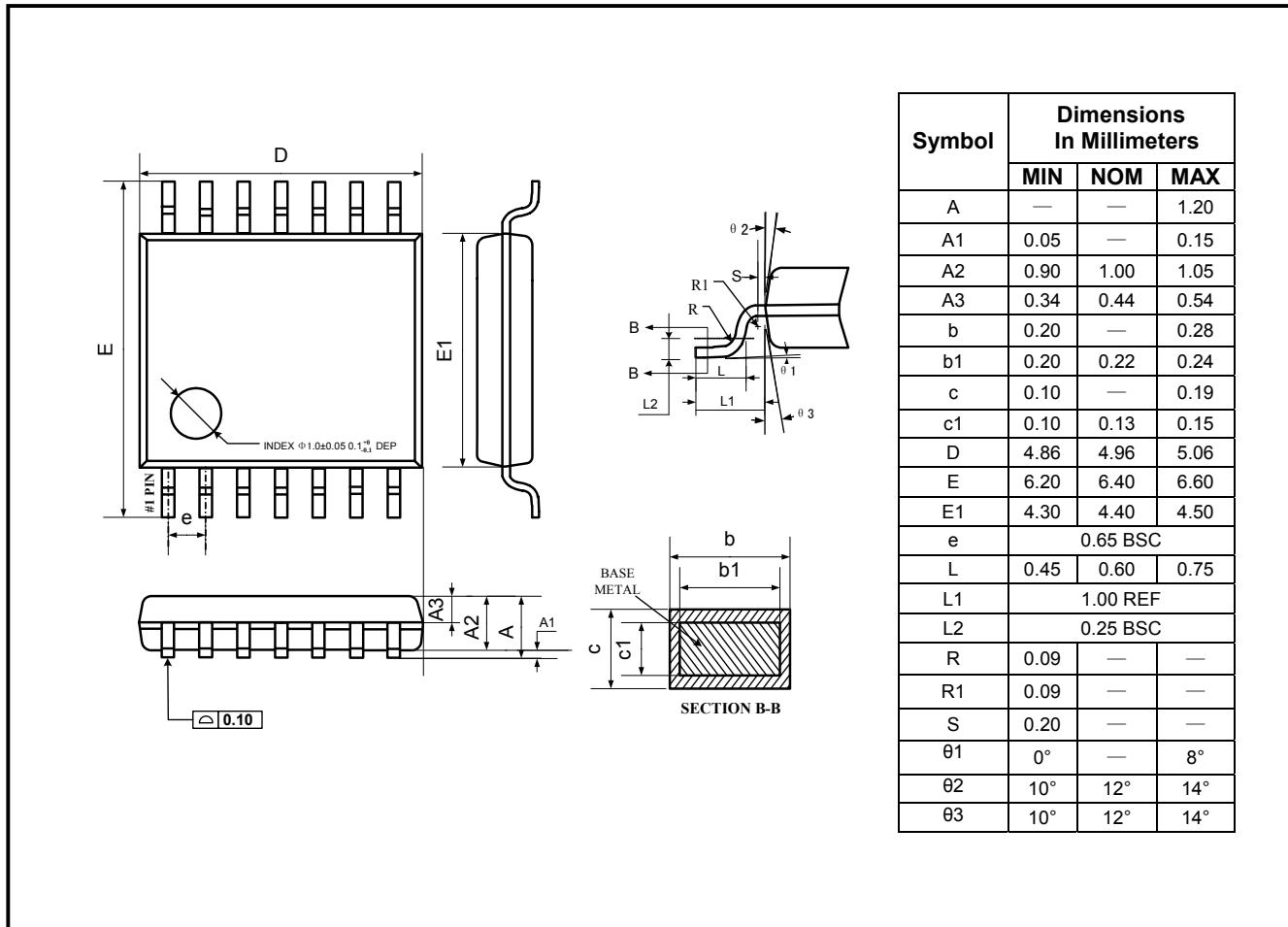
PACKAGE OUTLINE DIMENSIONS

SO-14



PACKAGE OUTLINE DIMENSIONS

TSSOP-14



03/2009 REV. B. 2

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