TL1454, TL1454Y
DUAL-CHANNEL PULSE-WIDTH-MODULATION (PWM) CONTROL CIRCUIT

- **Two Complete PWM Control Circuits**
- **Outputs Drive MOSFETs Directly**
- **Oscillator Frequency . . . 50 kHz to 2 MHz**
- **3.6-V to 20-V Supply-Voltage Range**
- **Low Supply Current . . . 3.5 mA Typ**
- **Adjustable Dead-Time Control, 0% to 100%**
- **1.25-V Reference**

D, N OR PW PACKAGE
(TOP VIEW)

CT	1	16	REF
RT	2	15	SCP
DTC1	3	14	DTC2
IN1+	4	13	IN2+
IN1–	5	12	IN2–
COMP1	6	11	COMP2
GND	7	10	V_{CC}
OUT1	8	9	OUT2
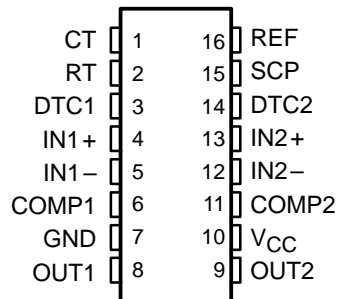
description

The TL1454 is a dual-channel pulse-width-modulation (PWM) control circuit, primarily intended for low-power, dc/dc converters. Applications include LCD displays, backlight inverters, notebook computers, and other products requiring small, high-frequency, dc/dc converters. Each PWM channel has its own error amplifier, PWM comparator, dead-time control comparator, and MOSFET driver. The voltage reference, oscillator, undervoltage lockout, and short-circuit protection are common to both channels.

Channel 1 is configured to drive n-channel MOSFETs in step-up or flyback converters, and channel 2 is configured to drive p-channel MOSFETs in step-down or inverting converters. The operating frequency is set with an external resistor and an external capacitor, and dead time is continuously adjustable from 0 to 100% duty cycle with a resistive divider network. Soft start can be implemented by adding a capacitor to the dead-time control (DTC) network. The error-amplifier common-mode input range includes ground, which allows the TL1454 to be used in ground-sensing battery chargers as well as voltage converters.

AVAILABLE OPTIONS

|	PACKAGED DEVICES†			CHIP FORM (Y)
T_A	SMALL OUTLINE (D)	PLASTIC DIP (N)	TSSOP (PW)	
−20°C to 85°C	TL1454CD	TL1454CN	TL1454CPWLE	TL1454Y
−40°C to 85°C	TL1454ID	TL1454IN	—	—

† The D package is available taped and reeled. Add the suffix R to the device name (e.g., TL1454CDR). The PW package is available only left-end taped and reeled (indicated by the LE suffix on the device type; e.g., TL1454CPWLE).

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TEXAS INSTRUMENTS

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functional block diagram
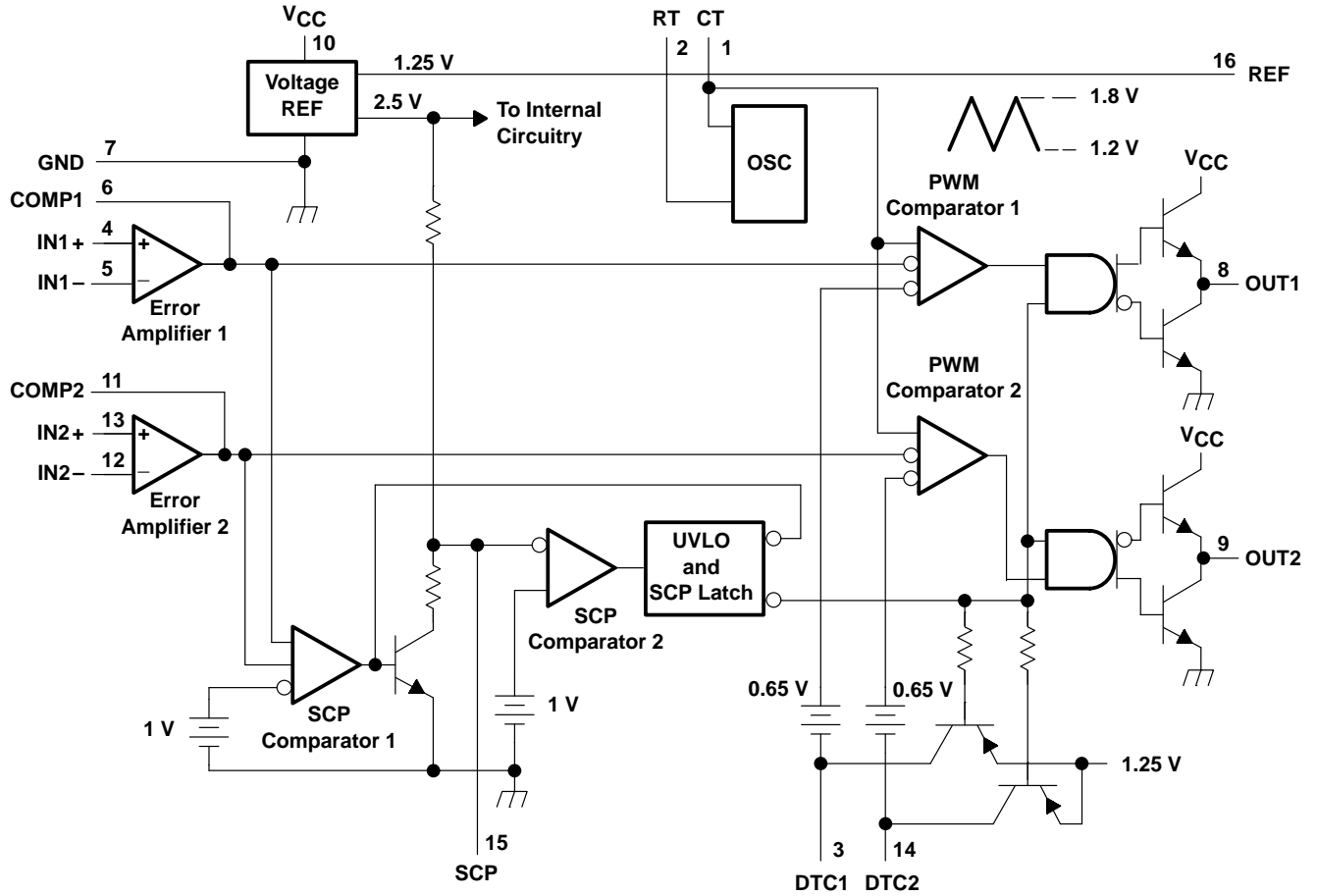
TL1454Y chip information

This chip, when properly assembled, displays characteristics similar to the TL1454C. Thermal compression or ultrasonic bonding may be used on the doped aluminum bonding pads. The chips may be mounted with conductive epoxy or a gold-silicon preform.
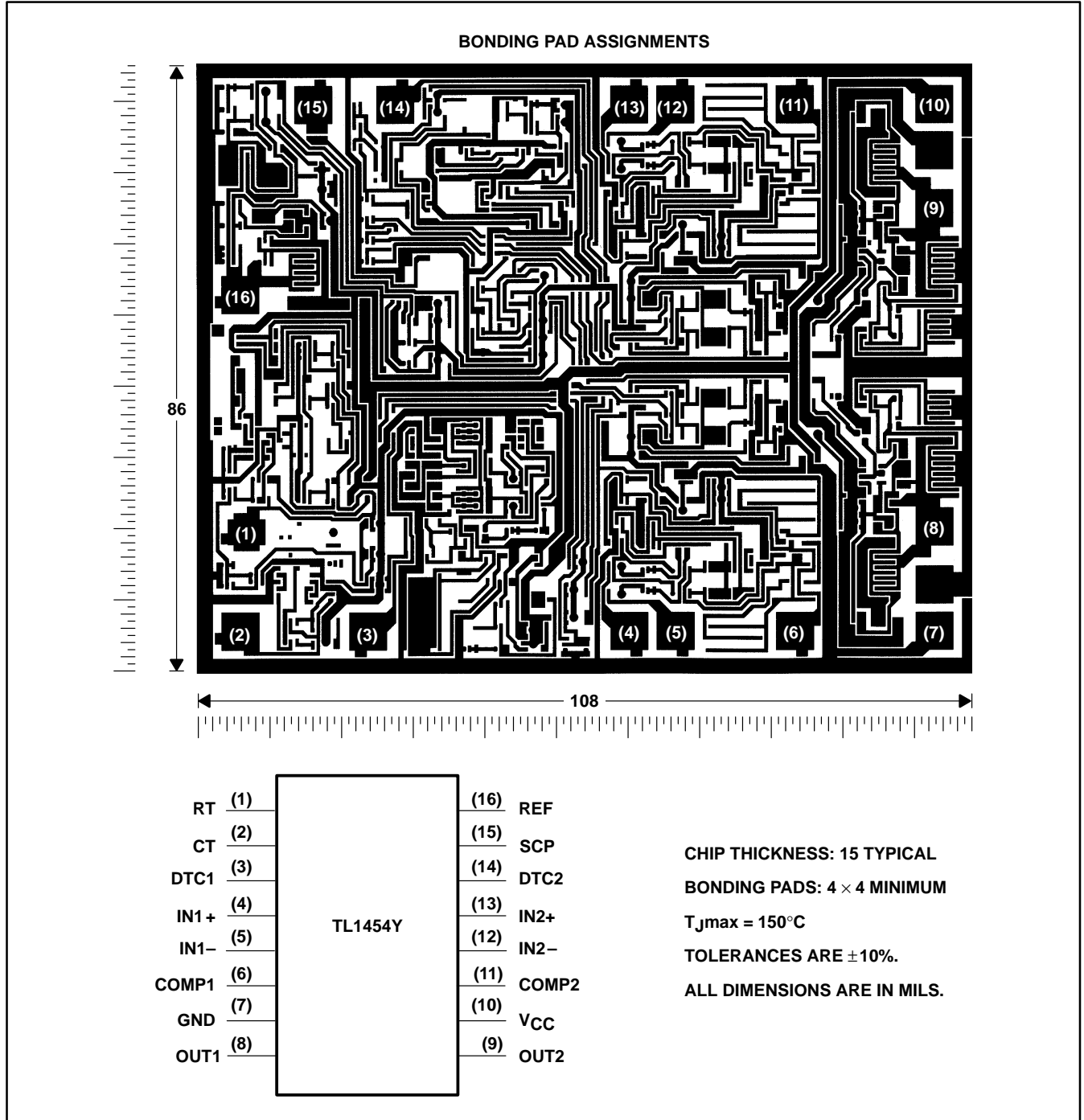
BONDING PAD ASSIGNMENTS

RT	(1)
CT	(2)
DTC1	(3)
IN1+	(4)
IN1–	(5)
COMP1	(6)
GND	(7)
OUT1	(8)

TL1454Y

(16)	REF
(15)	SCP
(14)	DTC2
(13)	IN2+
(12)	IN2–
(11)	COMP2
(10)	V$_{CC}$
(9)	OUT2

CHIP THICKNESS: 15 TYPICAL

BONDING PADS: 4 $\times$ 4 MINIMUM

T$_J$max = 150°C

TOLERANCES ARE ±10%.

ALL DIMENSIONS ARE IN MILS.

theory of operation

reference voltage

A linear regulator operating from V_{CC} generates a 2.5-V supply for the internal circuits and the 1.25-V reference, which can source a maximum of 1 mA for external loads. A small ceramic capacitor (0.047 μF to 0.1 μF) between REF and ground is recommended to minimize noise pickup.

error amplifier

The error amplifier generates the error signal used by the PWM to adjust the power-switch duty cycle for the desired converter output voltage. The signal is generated by comparing a sample of the output voltage to the voltage reference and amplifying the difference. An external resistive divider connected between the converter output and ground, as shown in Figure 1, is generally required to obtain the output voltage sample.

The amplifier output is brought out on COMP to allow the frequency response of the amplifier to be shaped with an external RC network to stabilize the feedback loop of the converter. DC loading on the COMP output is limited to 45 μA (the maximum amplifier source current capability).

Figure 1 illustrates the sense-divider network and error-amplifier connections for converters with positive output voltages. The divider network is connected to the noninverting amplifier input because the PWM has a phase inversion; the duty cycle decreases as the error-amplifier output increases.
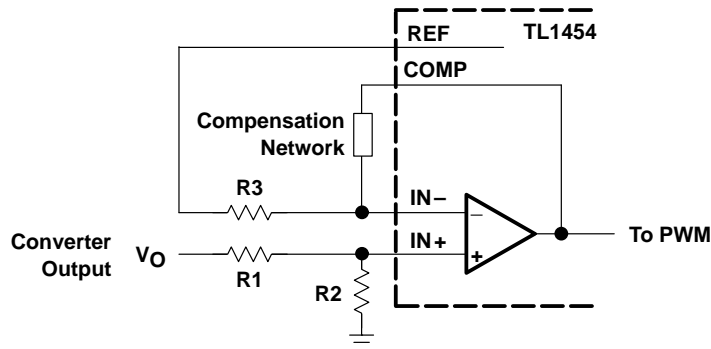
**Figure 1. Sense Divider/Error Amplifier
Configuration for Converters with Positive Outputs**

The output voltage is given by:

$$V_O = V_{ref} \left(1 + \frac{R1}{R2} \right)$$

where V_{ref} = 1.25 V.

The dc source resistance of the error-amplifier inputs should be 10 kΩ or less and approximately matched to minimize output voltage errors caused by the input-bias current. A simple procedure for determining appropriate values for the resistors is to choose a convenient value for R3 (10 kΩ or less) and calculate R1 and R2 using:

$$R_1 = \frac{R_3 V_O}{V_O - V_{ref}}$$

$$R_2 = \frac{R_3 V_O}{V_{ref}}$$

error amplifier

R1 and R2 should be tight-tolerance (±1% or better) devices with low and/or matched temperature coefficients to minimize output voltage errors. A device with a ±5% tolerance is suitable for R3.
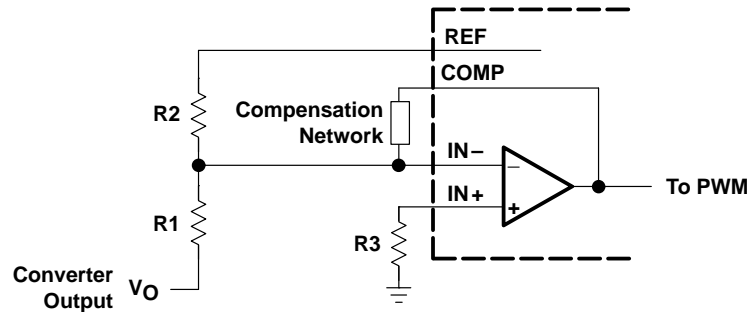
Figure 2. Sense Divider/Error Amplifier Configuration for Converters with Negative Outputs

Figure 2 shows the divider network and error-amplifier configuration for negative output voltages. In general, the comments for positive output voltages also apply for negative outputs. The output voltage is given by:

$$V_O \ = \ - \ \frac{R_1 V_{ref}}{R_2}$$

The design procedure for choosing the resistor value is to select a convenient value for R2 (instead of R3 in the procedure for positive outputs) and calculate R1 and R3 using:

$$R_1 \ = \ - \ \frac{R_2 V_O}{V_{ref}}$$

$$R_3 \ = \ \frac{R_1 R_2}{R_1 + R_2}$$

Values in the 10-kΩ to 20-kΩ range work well for R2. R3 can be omitted and the noninverting amplifier connected to ground in applications where the output voltage tolerance is not critical.

oscillator

The oscillator frequency can be set between 50 kHz and 2 MHz with a resistor connected between RT and GND and a capacitor between CT and GND (see Figure 3). Figure 6 is used to determine R_T and C_T for the desired operating frequency. Both components should be tight-tolerance, temperature-stable devices to minimize frequency deviation. A 1% metal-film resistor is recommended for R_T, and a 10%, or better, NPO ceramic capacitor is recommended for C_T.
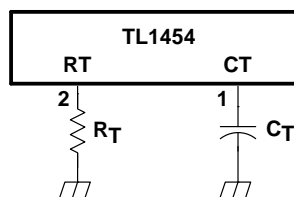
Figure 3. Oscillator Timing

dead-time control (DTC) and soft start

The two PWM channels have independent dead-time control inputs so that the maximum power-switch duty cycles can be limited to less then 100%. The dead-time is set with a voltage applied to DTC; the voltage is typically obtained from a resistive divider connected between the reference and ground as shown in Figure 4. Soft start is implemented by adding a capacitor between REF and DTC.

The voltage, V_{DT}, required to limit the duty cycle to a maximum value is given by

$$V_{DT} = V_{O(max)} - D\left(V_{O(max)} - V_{O(min)}\right) - 0.65$$

where $V_{O(max)}$ and $V_{O(min)}$ are obtained from Figure 9, and D is the maximum duty cycle.

Predicting the regulator startup or rise time is complicated because it depends on many variables, including: input voltage, output voltage, filter values, converter topology, and operating frequency. In general, the output will be in regulation within two time constants of the soft-start circuit. A five-to-ten millisecond time constant usually works well for low-power converters.

The DTC input can be grounded in applications where achieving a 100% duty cycle is desirable, such as a buck converter with a very low input-to-output differential voltge. However, grounding DTC prevents the inplementation of soft start, and the output voltage overshoot at power-on is likely to be very large. A better arrangement is to omit R_{DT1} (see Figure 4) and choose $R_{DT2} = 47$ kΩ. This configuration ensures that the duty cycle can reach 100% and still allows the designer to implement soft start using C_{SS}.
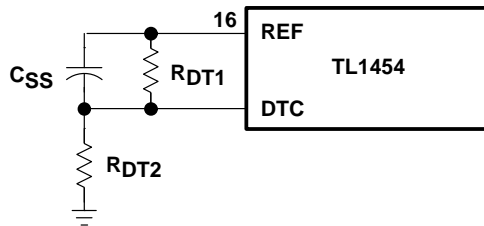

Figure 4. Dead-Time Control and Soft Start

PWM comparator

Each of the PWM comparators has dual inverting inputs. One inverting input is connected to the output of the error amplifier; the other inverting input is connected to the DTC terminal. Under normal operating conditions, when either the error-amplifier output or the dead-time control voltage is higher than that for the PWM triangle wave, the output stage is set inactive (OUT1 low and OUT2 high), turning the external power stage off.

undervoltage-lockout (UVLO) protection

The undervoltage-lockout circuit turns the output circuit off and resets the SCP latch whenever the supply voltage drops too low (to approximately 2.9 V) for proper operation. A hysteresis voltage of 200 mV eliminates false triggering on noise and chattering.

short-circuit protection (SCP)

The TL1454 SCP function prevents damage to the power switches when the converter output is shorted to ground. In normal operation, SCP comparator 1 clamps SCP to approximately 185 mV. When one of the converter outputs is shorted, the error amplifier output (COMP) will be driven below 1 V to maximize duty cycle and force the converter output back up. When the error amplifier output drops below 1 V, SCP comparator 1 releases SCP, and capacitor, C_{SCP}, which is connected between SCP and GND, begins charging. If the error-amplifier output rises above 1 V before C_{SCP} is charged to 1 V, SCP comparator 1 discharges C_{SCP} and normal operation resumes. If C_{SCP} reaches 1 V, SCP comparator 2 turns on and sets the SCP latch, which turns off the output drives and resets the soft-start circuit. The latch remains set until the supply voltage is lowered to 2 V or less, or C_{SCP} is discharged externally.

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short-circuit protection (SCP) (continued)

The SCP time-out period must be greater than the converter start-up time or the converter will not start. Because high-value capacitor tolerances tend to be ±20% or more and IC resistor tolerances are loose as well, it is best to choose an SCP time-out period ten-to-fifteen times greater than the converter startup time. The value of C_{SCP} may be determined using Figure 6, or it can be calculated using:

$$C_{SCP} = \frac{T_{SCP}}{80.3}$$

where C_{SCP} is in µF and T_{SCP} is the time-out period in ms.

output stage

The output stage of the TL1454 is a totem-pole output with a maximum source/sink current rating of 40 mA and a voltage rating of 20 V. The output is controlled by a complementary output AND gate and is turned on (sourcing current for OUT1, sinking current for OUT2) when all the following conditions are met: 1) the oscillator triangle wave voltage is higher than both the DTC voltage and the error-amplifier output voltage, 2) the undervoltage-lockout circuit is inactive, and 3) the short-circuit protection circuit is inactive.

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†

Supply voltage, V_{CC} (see Note 1) . . . . 23 V
Error amplifier input voltage: IN1+, IN1−, IN2+, IN2− . . . . 23 V
Output voltage: OUT1, OUT2 . . . . 20 V
Continuous output current: OUT1, OUT2 . . . . ±200 mA
Peak output current: OUT1, OUT2 . . . . 1 A
Continuous total dissipation . . . . See Dissipation Rating Table
Operating free-air temperature range, T_A: C suffix . . . . −20°C to 85°C
                                                                I suffix . . . . −40°C to 85°C
Storage temperature range, T_{stg} . . . . −65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds . . . . 260°C

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

NOTE 1: All voltage values are with respect to network GND.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25°C$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25°C$	$T_A = 70°C$ POWER RATING	$T_A = 85°C$ POWER RATING
D	950 mW	7.6 mW/°C	608 mW	494 mW
N	1250 mW	10.0 mW/°C	800 mW	650 mW
PW	500 mW	4.0 mW/°C	320 mW	260 mW

recommended operating conditions

		MIN	MAX	UNIT
Supply voltage, V_{CC}		3.6	20	V
Error amplifier common-mode input voltage		−0.2	1.45	V
Output voltage, V_O			20	V
Output current, I_O			±40	mA
COMP source current			−45	μA
COMP sink current			100	μA
Reference output current			1	mA
COMP dc load resistance		100		kΩ
Timing capacitor, C_T		10	4000	pF
Timing resistor, R_T		5.1	100	kΩ
Oscillator frequency		50	2000	kHz
Operating free-air temperature, T_A	TL1454C	−20	85	°C
	TL1454I	−40	85	

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electrical characteristics over recommended operating free-air temperature range, V_{CC} = 6 V, f_{osc} = 500 kHz (unless otherwise noted)

reference

PARAMETER		TEST CONDITIONS		TL1454			UNIT
				MIN	TYP	MAX	
V_{ref}	Output voltage, REF	I_O = 1 mA,	T_A = 25°C	1.23	1.25	1.28	V
		I_O = 1 mA		1.2		1.31	
	Input regulation	V_{OC} = 3.6 V to 20 V,	I_O = 1 mA		2	6	mV
	Output regulation	I_O = 0.1 mA to 1 mA			1	7.5	mV
	Output voltage change with temperature	T_A = $T_{A(min)}$ to 25°C,	I_O = 1 mA	−12.5	−1.25	12.5	mV
		T_A = 25°C to 85°C,	I_O = 1 mA	−12.5	−2.5	12.5	
I_{OS}	Short-circuit output current	V_{ref} = 0 V			30		mA

undervoltage lockout (UVLO)

PARAMETER		TEST CONDITIONS	TL1454			UNIT
			MIN	TYP	MAX	
V_{IT+}	Positive-going threshold voltage			2.9		V
V_{IT-}	Negative-going threshold voltage	T_A = 25°C		2.7		V
V_{hys}	Hysteresis, V_{IT+} − V_{IT-}		100	200		mV

short-circuit protection (SCP)

PARAMETER		TEST CONDITIONS		TL1454			UNIT
				MIN	TYP	MAX	
V_{IT}	Input threshold voltage	T_A = 25°C		0.95	1	1.05	V
V_{stby}†	Standby voltage	No pullup		140	185	230	mV
$V_{I(latched)}$	Latched-mode input voltage				60	120	mV
$V_{IT(COMP)}$	Comparator threshold voltage	COMP1, COMP2			1		V
	Input source current	T_A = 25°C,	$V_{O(SCP)}$ = 0	−5	−15	−20	µA

† This symbol is not presently listed within EIA / JEDEC standards for semiconductor symbology.

oscillator

PARAMETER		TEST CONDITIONS		TL1454			UNIT
				MIN	TYP	MAX	
f_{osc}	Frequency	C_T = 120 pF,	R_T = 10 kΩ		500		kHz
	Standard deviation of frequency				50		kHz
	Frequency change with voltage	V_{CC} = 3.6 V to 20 V,	T_A = 25°C		5		kHz
	Frequency change with temperature	T_A = $T_{A(min)}$ to 25°C			−2	±20	kHz
		T_A = 25°C to 85°C			−10	±20	
	Maximum ramp voltage				1.8		V
	Minimum ramp voltage				1.1		V

electrical characteristics over recommended operating free-air temperature range, V_{CC} = 6 V, f_{osc} = 500 kHz (unless otherwise noted) (continued)

dead-time control (DTC)

PARAMETER		TEST CONDITIONS	TL1454			UNIT
			MIN	TYP	MAX	
V_{IT}	Input threshold voltage	Duty cycle = 0%	1	1.1	1.2	V
		Duty cycle = 100%	0.4	0.5	0.6	
$V_{I(latched)}$	Latched-mode input voltage			1.2		V
I_{IB}	Common-mode input bias current	DTC1, IN1+ ≈ 1.2 V			4	μA
	Latched-mode (source) current	T_A = 25°C		−100		μA

error-amplifier

PARAMETER		TEST CONDITIONS	TL1454			UNIT
			MIN	TYP	MAX	
V_{IO}	Input offset voltage				6	mV
I_{IO}	Input offset current	V_O = 1.25 V, V_{IC} = 1.25 V			100	nA
I_{IB}	Input bias current			−160	−500	nA
V_{ICR}	Input voltage range	V_{CC} = 3.6 V to 20 V	−0.2 to 1.45			V
A_V	Open-loop voltage gain	R_{FB} = 200 kΩ	70	80		dB
	Unity-gain bandwidth			3		MHz
CMRR	Common-mode rejection ratio		60	80		dB
$V_{OM(max)}$	Positive output voltage swing		2.3	2.43		V
$V_{OM(min)}$	Negative output voltage swing			0.63	0.8	
I_{O+}	Output sink current	V_{ID} = −0.1 V, V_O = 1.20 V	0.1	0.5		mA
I_{O-}	Output source current	V_{ID} = 0.1 V, V_O = 1.80 V	−45	−70		μA

output

PARAMETER		TEST CONDITIONS	TL1454			UNIT
			MIN	TYP	MAX	
V_{OH}	High-level output voltage	I_O = −8 mA	V_{CC}−2	4.5		V
		I_O = −40 mA	V_{CC}−2	4.4		
V_{OL}	Low-level output voltage	I_O = 8 mA		0.1	0.4	V
		I_O = 40 mA		1.8	2.5	
t_{rv}	Output voltage rise time	C_L = 2000 pF, T_A = 25°C		220		ns
t_{fv}	Output voltage fall time			220		

supply current

PARAMETER		TEST CONDITIONS	TL1454			UNIT
			MIN	TYP	MAX	
$I_{CC(stdby)}$	Standby supply current	RT open, CT = 1.5 V, V_O (COMP1, COMP2) = 1.25 V, No load		3.1	6	mA
$I_{CC(average)}$	Average supply current	R_T = 10 kΩ, C_T = 120 pF, 50% duty cycle, Outputs open		3.5	7	mA

electrical characteristics, V_{CC} = 6 V, f_{osc} = 500 kHz, T_A = 25°C (unless otherwise noted)

reference

PARAMETER		TEST CONDITIONS	TL1454Y			UNIT
			MIN	TYP	MAX	
V_{ref}	Output voltage, REF	I_O = 1 mA,		1.25		V
	Input regulation	V_{OC} = 3.6 V to 20 V, I_O = 1 mA		2		mV
	Output regulation	I_O = 0.1 mA to 1 mA		1		mV
	Output voltage change with temperature	I_O = 1 mA		−1.25		mV
		I_O = 1 mA		−2.5		
I_{OS}	Short-circuit output current	V_{ref} = 0 V		30		mA

undervoltage lockout (UVLO)

PARAMETER		TEST CONDITIONS	TL1454Y			UNIT
			MIN	TYP	MAX	
V_{IT+}	Positive-going threshold voltage			2.9		V
V_{IT-}	Negative-going threshold voltage			2.7		V
V_{hys}	Hysteresis, $V_{IT+} - V_{IT-}$			200		mV

short-circuit protection (SCP)

PARAMETER		TEST CONDITIONS	TL1454Y			UNIT
			MIN	TYP	MAX	
V_{IT}	Input threshold voltage			1		V
V_{stby}†	Standby voltage	No pullup		185		mV
$V_{I(latched)}$	Latched-mode input voltage			60		mV
$V_{IT(COMP)}$	Comparator threshold voltage	COMP1, COMP2		1		V
	Input source current	$V_{O(SCP)}$ = 0		−15		μA

† This symbol is not presently listed within EIA/JEDEC standards for semiconductor symbology.

oscillator

PARAMETER		TEST CONDITIONS	TL1454Y			UNIT
			MIN	TYP	MAX	
f_{osc}	Frequency	C_T = 120 pF, R_T = 10 kΩ		500		kHz
	Standard deviation of frequency			50		kHz
	Frequency change with voltage	V_{CC} = 3.6 V to 20 V		5		kHz
	Frequency change with temperature			−2		kHz
				−10		
	Maximum ramp voltage			1.8		V
	Minimum ramp voltage			1.1		V

electrical characteristics, V_{CC} = 6 V, f_{osc} = 500 kHz, T_A = 25°C (unless otherwise noted) (continued)

dead-time control (DTC)

PARAMETER		TEST CONDITIONS	TL1454Y			UNIT
			MIN	TYP	MAX	
V_{IT}	Input threshold voltage	Duty cycle = 0%		1.1		V
		Duty cycle = 100%		0.5		
$V_{I(latched)}$	Latched-mode input voltage			1.2		V
	Latched-mode (source) current			−100		µA

error-amplifier

PARAMETER		TEST CONDITIONS	TL1454Y			UNIT
			MIN	TYP	MAX	
I_{IB}	Input bias current	V_O = 1.25 V, V_{IC} = 1.25 V		−160		nA
A_V	Open-loop voltage gain	R_{FB} = 200 kΩ		80		dB
	Unity-gain bandwidth			3		MHz
CMRR	Common-mode rejection ratio			80		dB
$V_{OM(max)}$	Positive output voltage swing			2.43		V
$V_{OM(min)}$	Negative output voltage swing			0.63		
I_{O+}	Output sink current	V_{ID} = −0.1 V, V_O = 1.20 V		0.5		mA
I_{O-}	Output source current	V_{ID} = 0.1 V, V_O = 1.80 V		−70		µA

output

PARAMETER		TEST CONDITIONS	TL1454Y			UNIT
			MIN	TYP	MAX	
V_{OH}	High-level output voltage	I_O = −8 mA		4.5		V
		I_O = −40 mA		4.4		
V_{OL}	Low-level output voltage	I_O = 8 mA		0.1		V
		I_O = 40 mA		1.8		
t_{rv}	Output voltage rise time	C_L = 2000 pF		220		ns
t_{fv}	Output voltage fall time			220		

supply current

PARAMETER		TEST CONDITIONS	TL1454Y			UNIT
			MIN	TYP	MAX	
$I_{CC(stdby)}$	Standby supply current	RT open, CT = 1.5 V, V_O (COMP1, COMP2) = 1.25 V, No load		3.1		mA
$I_{CC(average)}$	Average supply current	R_T = 10 kΩ, C_T = 120 pF, 50% duty cycle, Outputs open		3.5		mA

PARAMETER MEASUREMENT INFORMATION
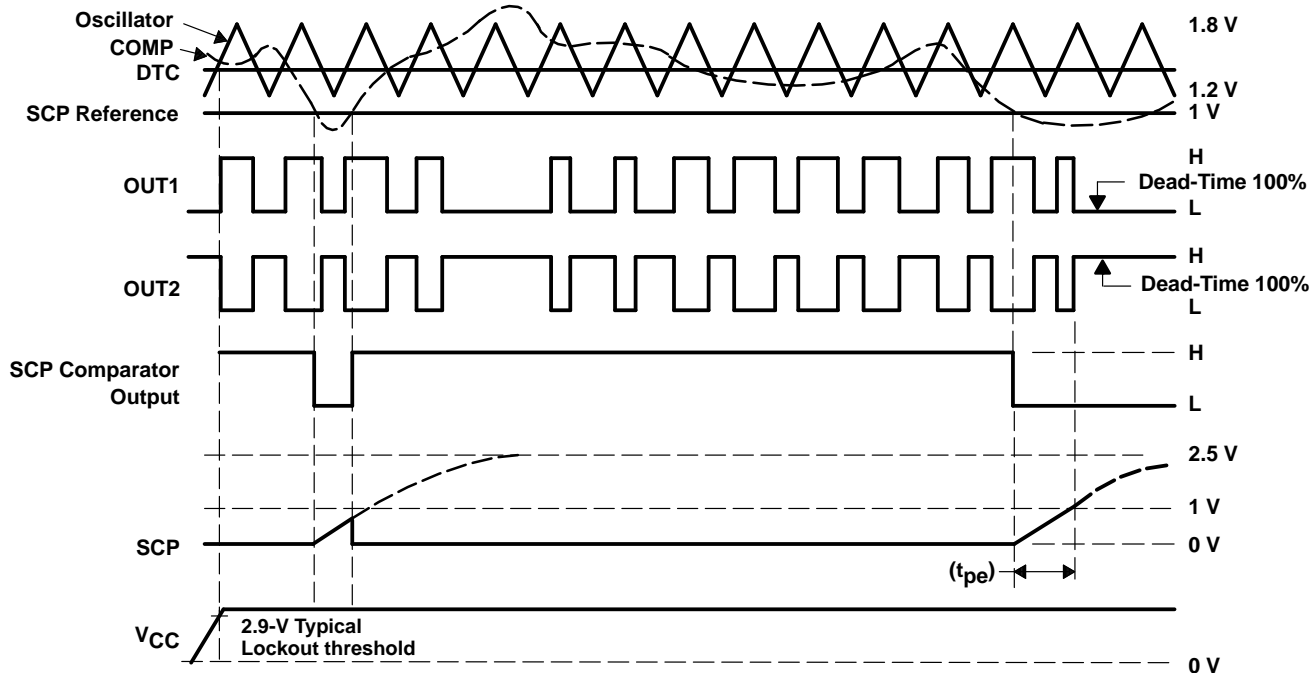
Figure 5. Timing Diagram

TYPICAL CHARACTERISTICS
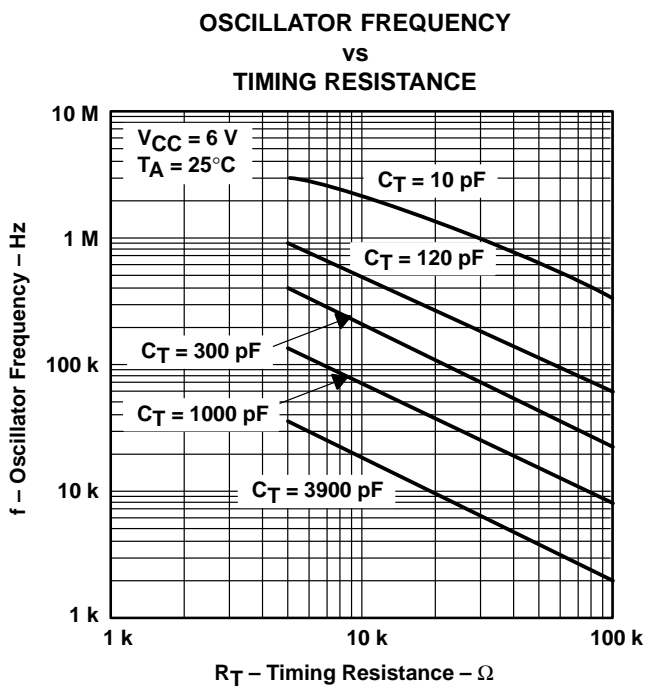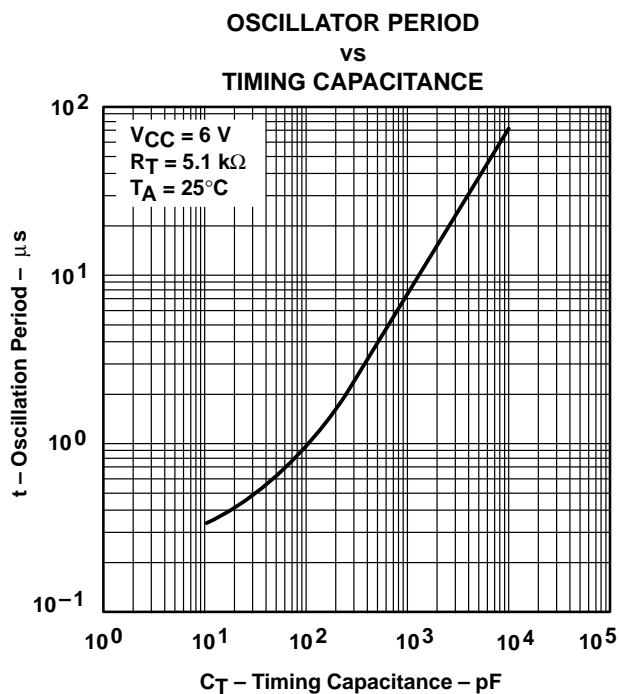
OSCILLATOR FREQUENCY
vs
TIMING RESISTANCE

Figure 6

OSCILLATOR PERIOD
vs
TIMING CAPACITANCE

Figure 7

TYPICAL CHARACTERISTICS

OSCILLATOR FREQUENCY
vs
FREE-AIR TEMPERATURE

Figure 8

PWM TRIANGLE WAVEFORM AMPLITUDE
vs
TIMING CAPACITANCE

Figure 9

DTC INPUT THRESHOLD VOLTAGE
vs
FREE-AIR TEMPERATURE

Figure 10

SCP TIME-OUT PERIOD
vs
SCP CAPACITANCE
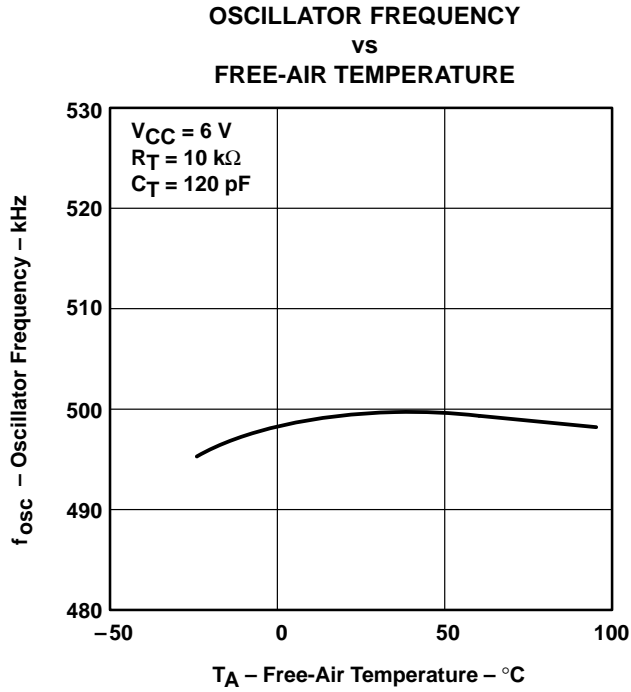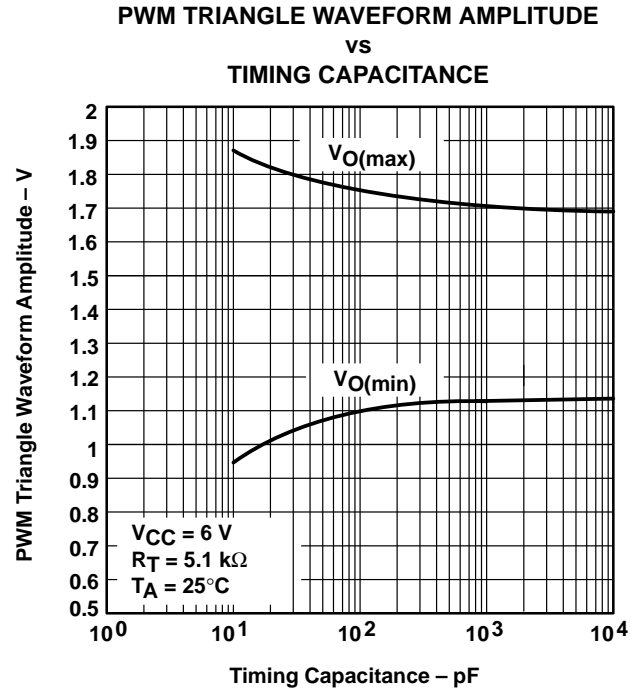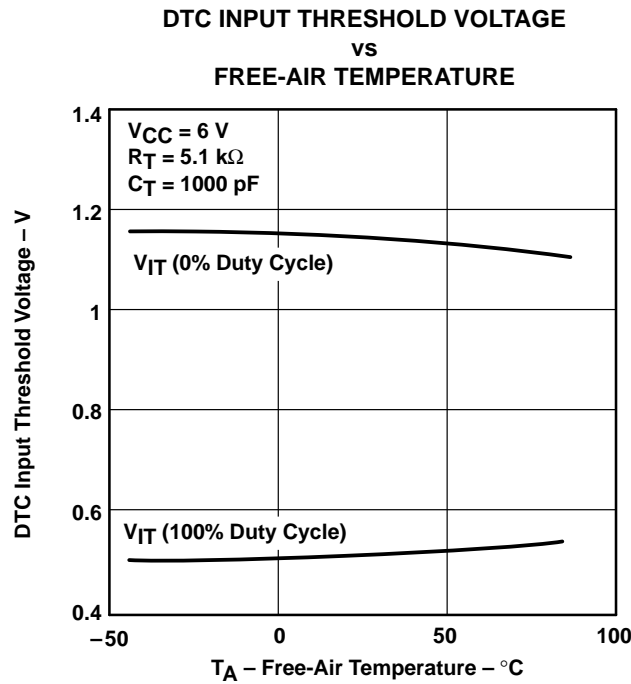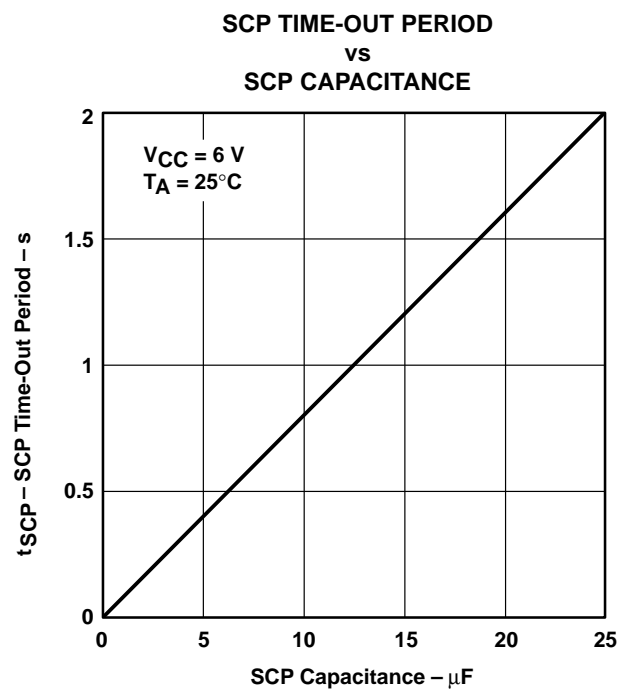
Figure 11

TYPICAL CHARACTERISTICS
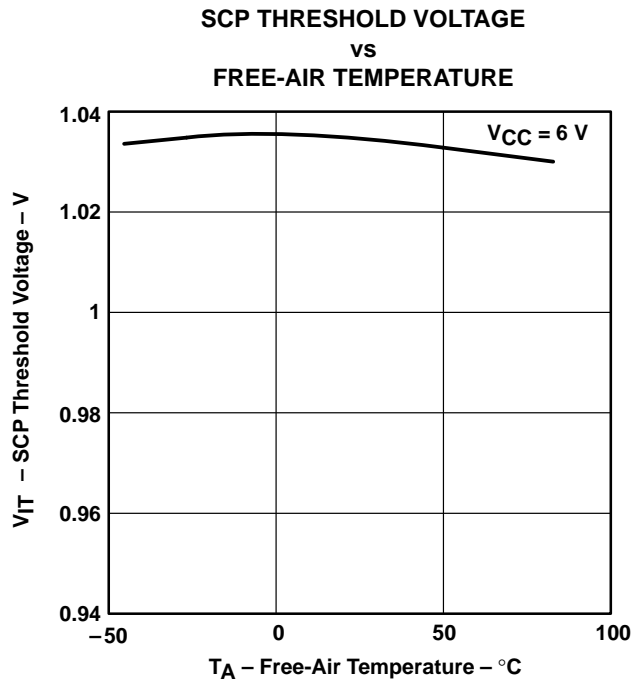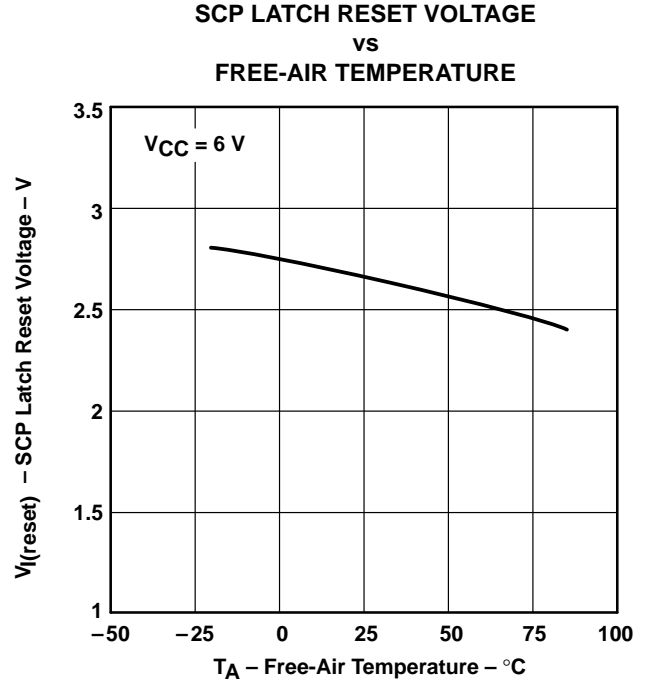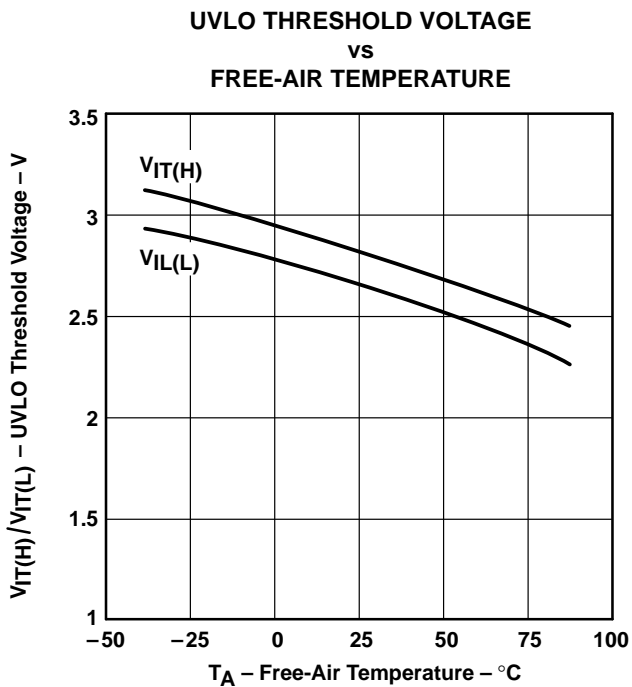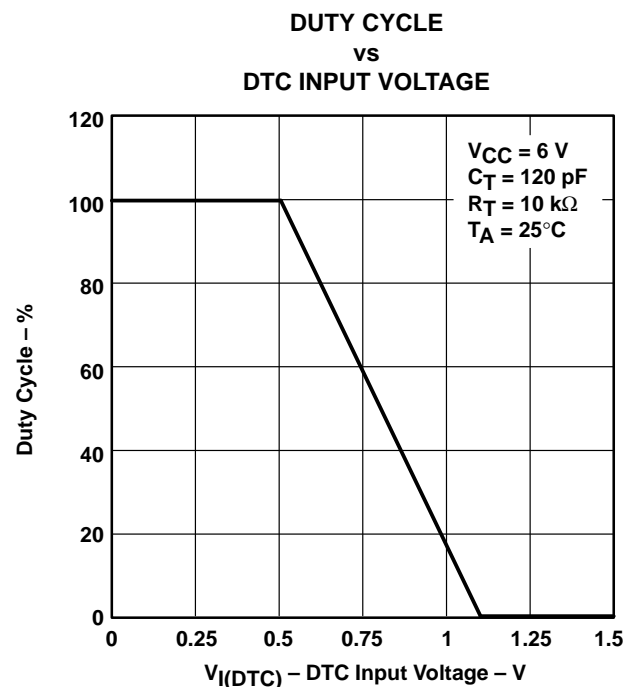
SCP THRESHOLD VOLTAGE
vs
FREE-AIR TEMPERATURE

Figure 12

SCP LATCH RESET VOLTAGE
vs
FREE-AIR TEMPERATURE

Figure 13

UVLO THRESHOLD VOLTAGE
vs
FREE-AIR TEMPERATURE

Figure 14

DUTY CYCLE
vs
DTC INPUT VOLTAGE

Figure 15

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TYPICAL CHARACTERISTICS

**ERROR-AMPLIFIER MAXIMUM OUTPUT VOLTAGE
vs
SOURCE CURRENT**

Figure 16

**ERROR-AMPLIFIER MINIMUM OUTPUT VOLTAGE
vs
SINK CURRENT**

Figure 17

**ERROR AMPLIFIER MAXIMUM
PEAK-TO-PEAK OUTPUT VOLTAGE SWING
vs
FREQUENCY**

Figure 18

**ERROR-AMPLIFIER MINIMUM OUTPUT
VOLTAGE SWING
vs
FREE-AIR TEMPERATURE**

Figure 19

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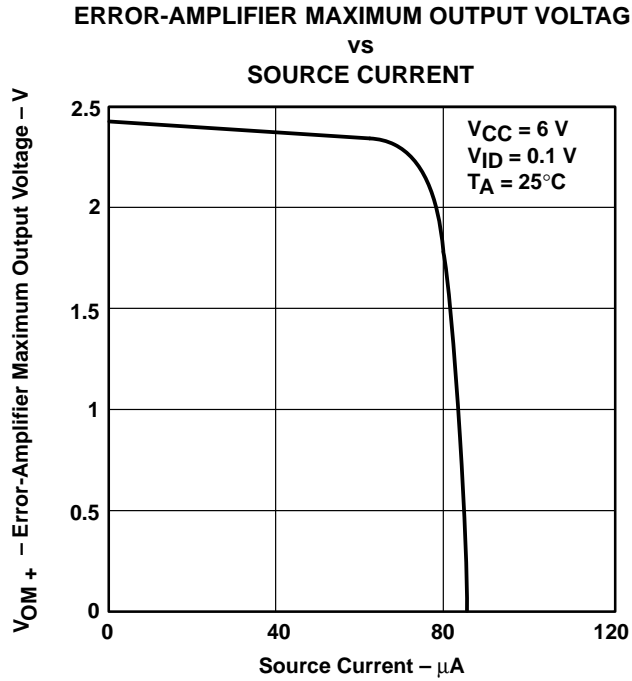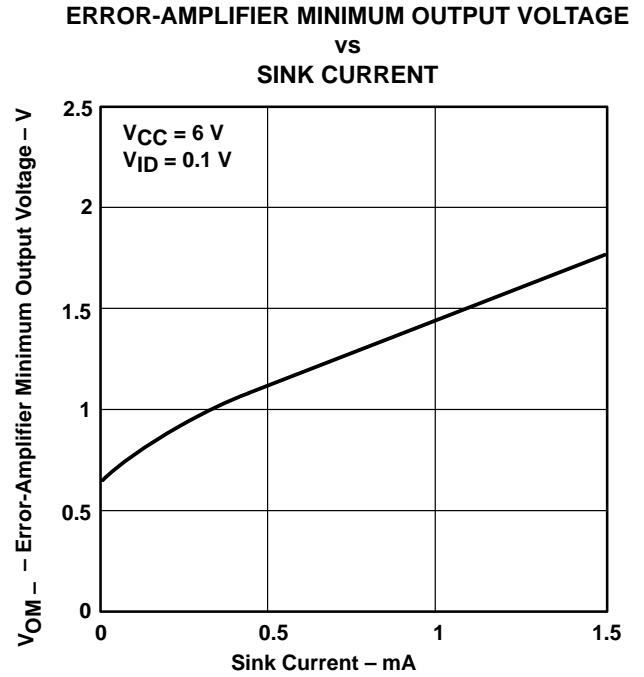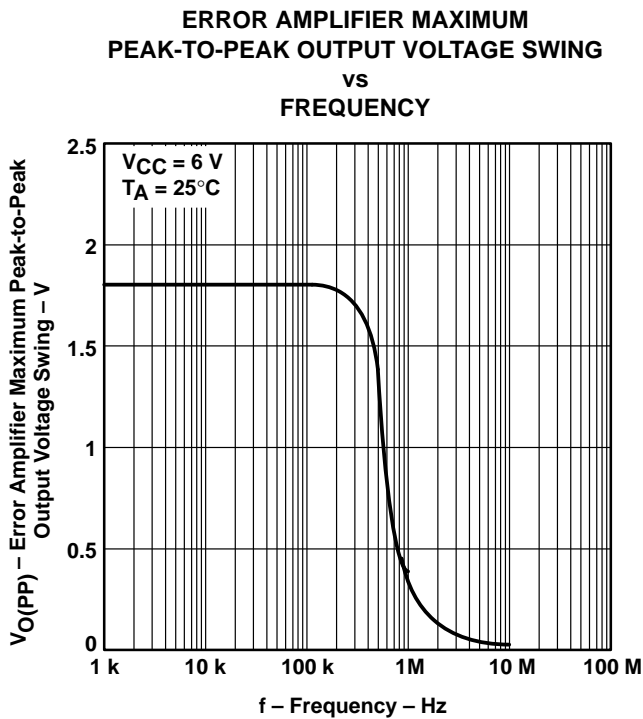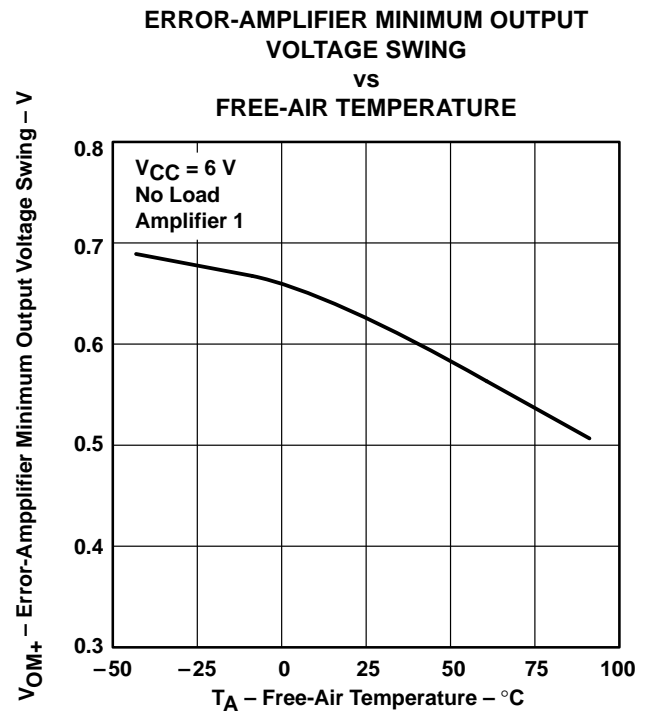
TYPICAL CHARACTERISTICS

ERROR AMPLIFIER OPEN-LOOP GAIN AND PHASE SHIFT
vs
FREQUENCY
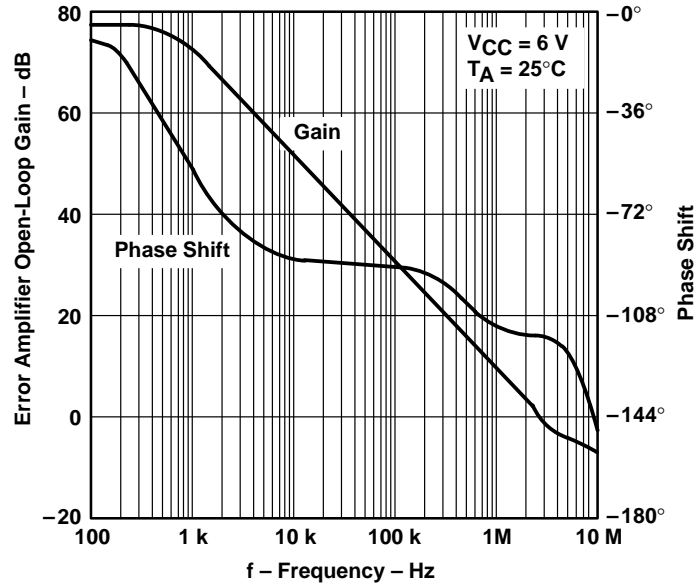
Figure 20

ERROR-AMPLIFIER POSITIVE OUTPUT
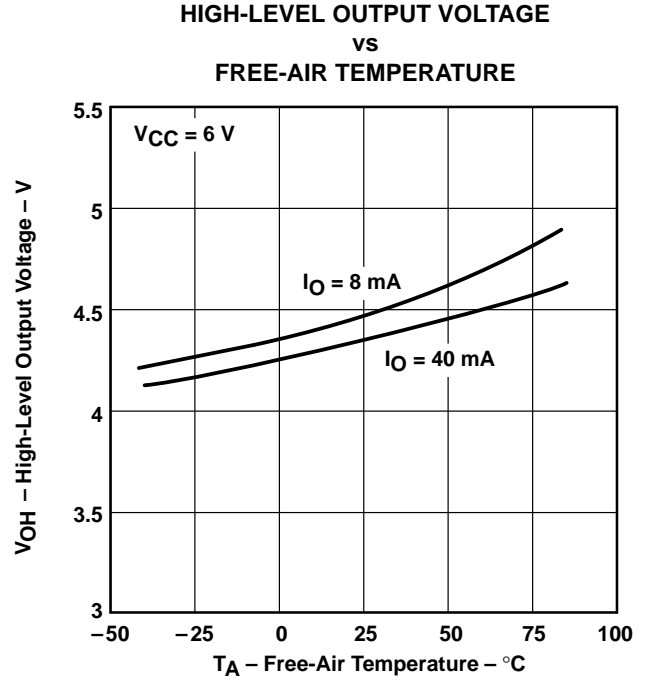VOLTAGE SWING
vs
FREE-AIR TEMPERATURE
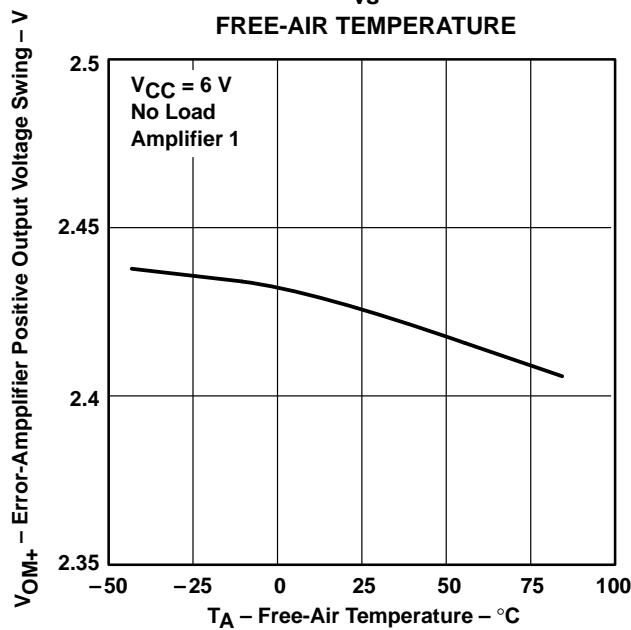
Figure 21

TYPICAL CHARACTERISTICS
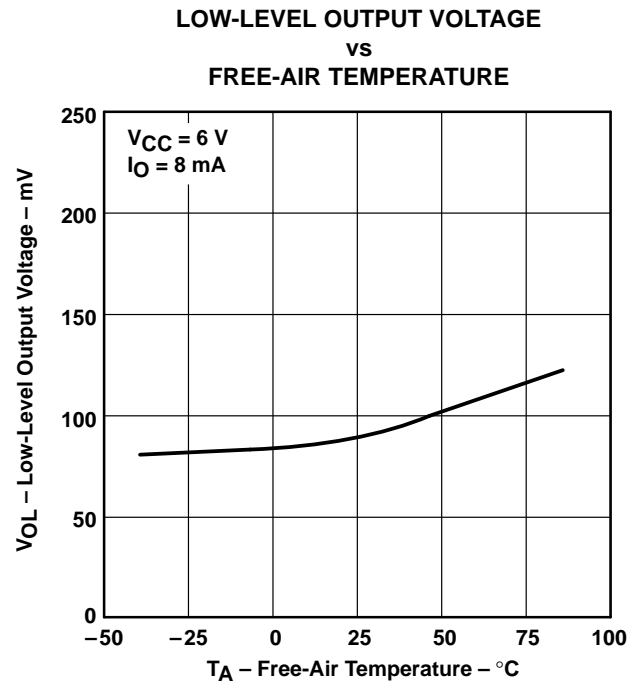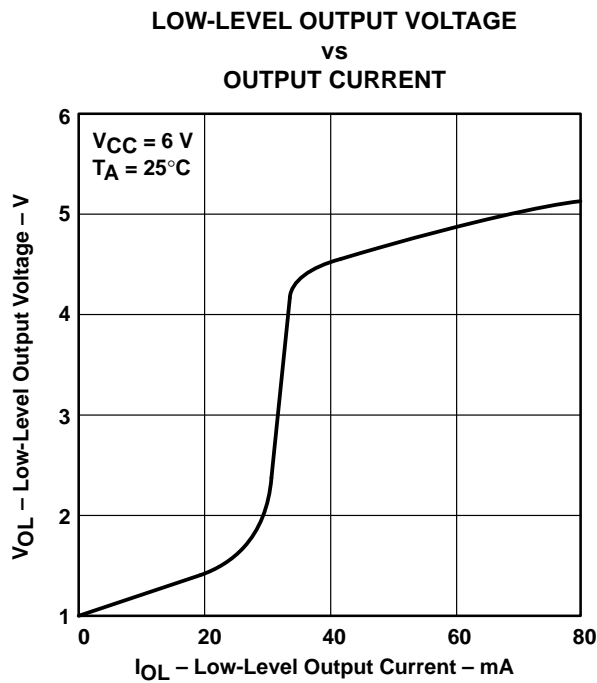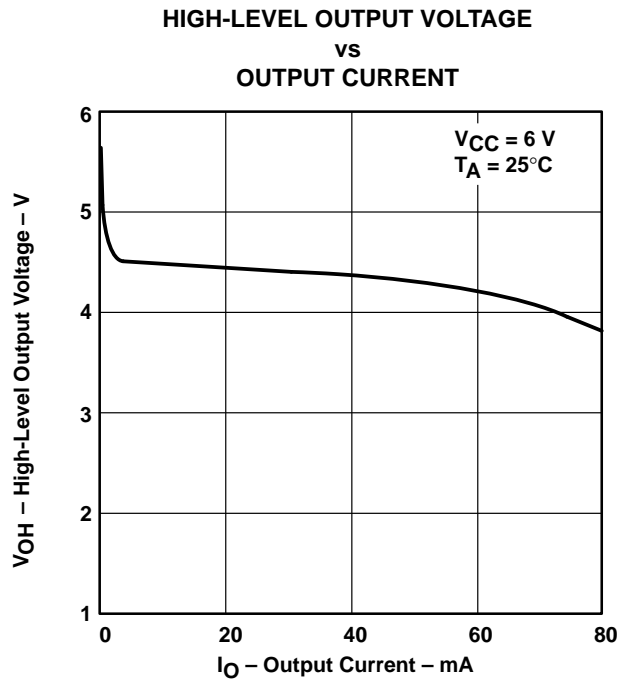
HIGH-LEVEL OUTPUT VOLTAGE
vs
OUTPUT CURRENT

Figure 22

HIGH-LEVEL OUTPUT VOLTAGE
vs
FREE-AIR TEMPERATURE

Figure 23

LOW-LEVEL OUTPUT VOLTAGE
vs
OUTPUT CURRENT

Figure 24

LOW-LEVEL OUTPUT VOLTAGE
vs
FREE-AIR TEMPERATURE

Figure 25

TYPICAL CHARACTERISTICS

LOW-LEVEL OUTPUT VOLTAGE
vs
FREE-AIR TEMPERATURE

Figure 26

AVERAGE SUPPLY CURRENT
vs
FREE-AIR TEMPERATURE

Figure 27

STANDBY SUPPLY CURRENT
vs
SUPPLY VOLTAGE

Figure 28

STANDBY SUPPLY CURRENT
vs
FREE-AIR TEMPERATURE

Figure 29

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TYPICAL CHARACTERISTICS

**REFERENCE VOLTAGE
vs
SUPPLY VOLTAGE**

$T_A = 25°C$

Figure 30

**REFERENCE VOLTAGE
vs
SUPPLY VOLTAGE**

$I_O = 1 mA$
$T_A = 25°C$

Figure 31

**REFERENCE VOLTAGE
vs
FREE-AIR TEMPERATURE**

$V_{CC} = 6 V$
$I_O = -1 mA$

Figure 32

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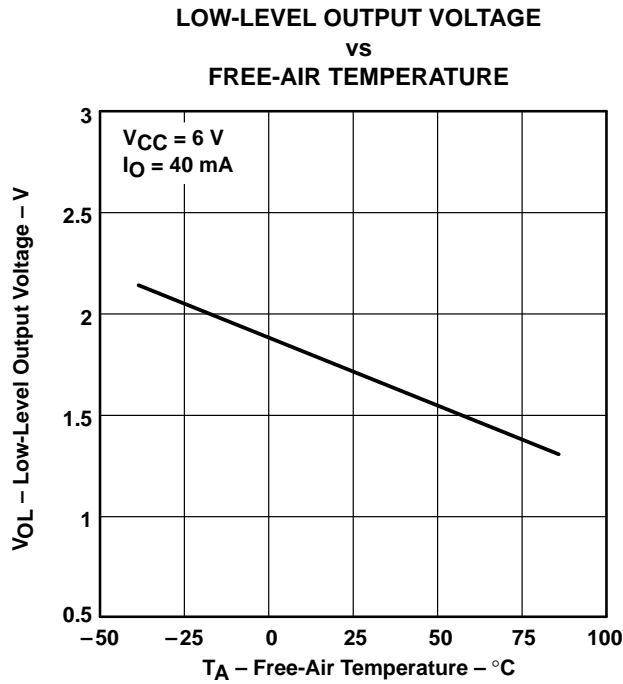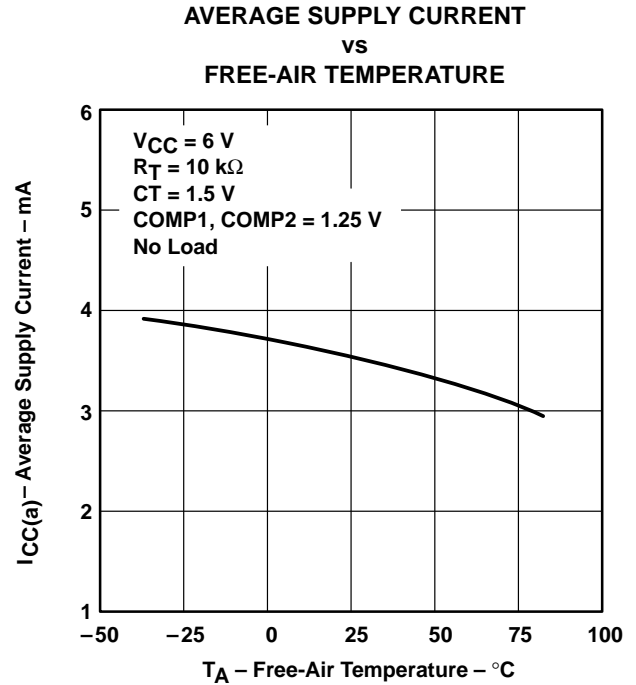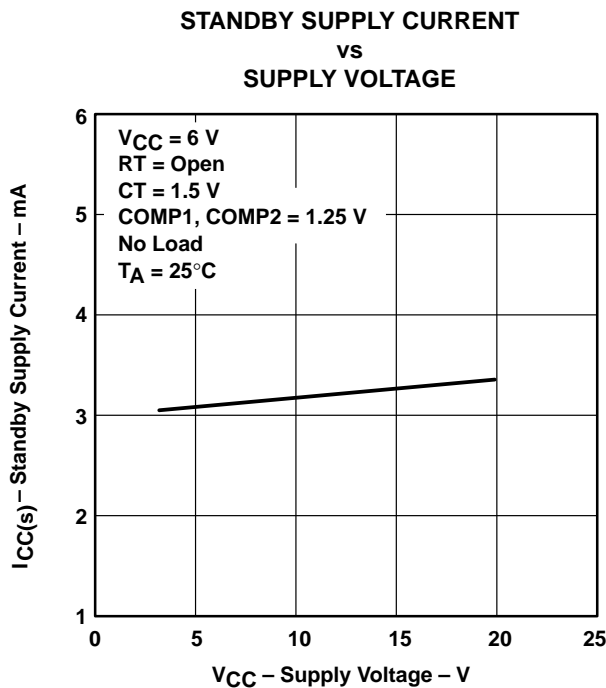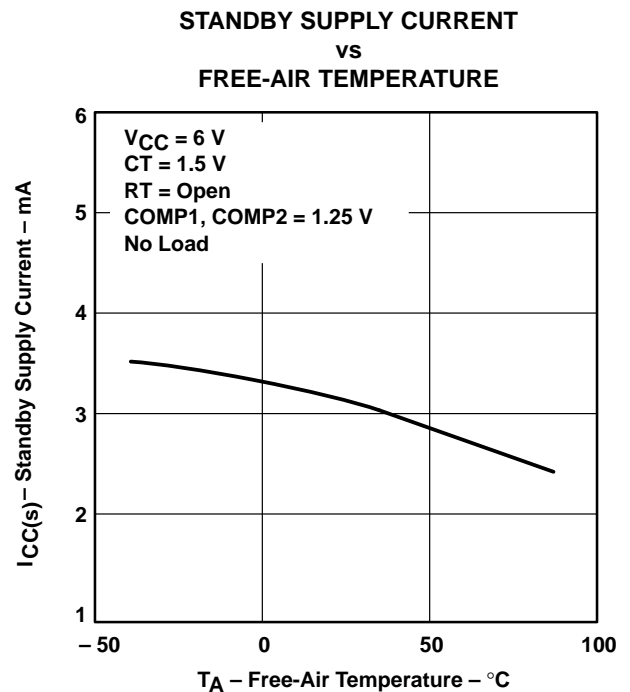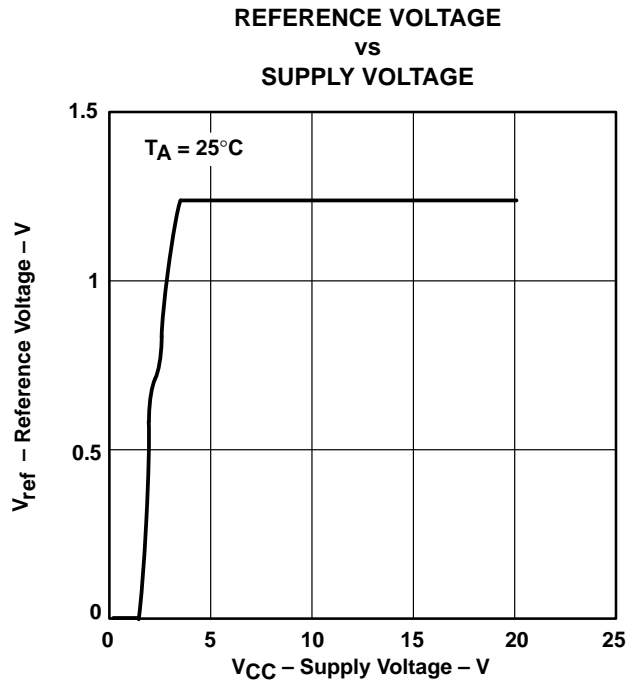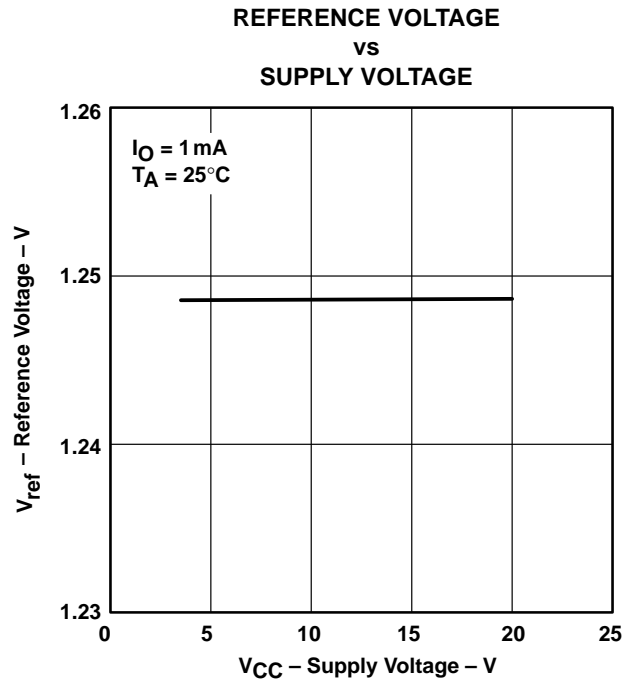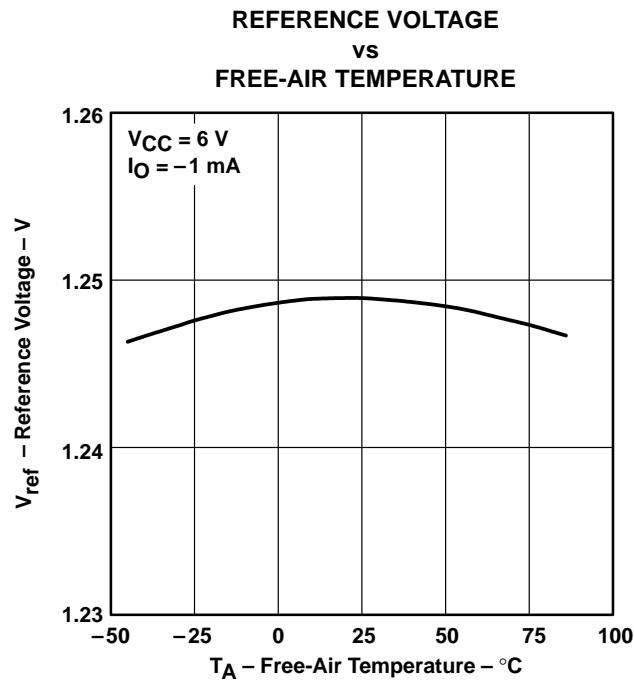
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MECHANICAL DATA

D (R-PDSO-G)**   **PLASTIC SMALL-OUTLINE PACKAGE**
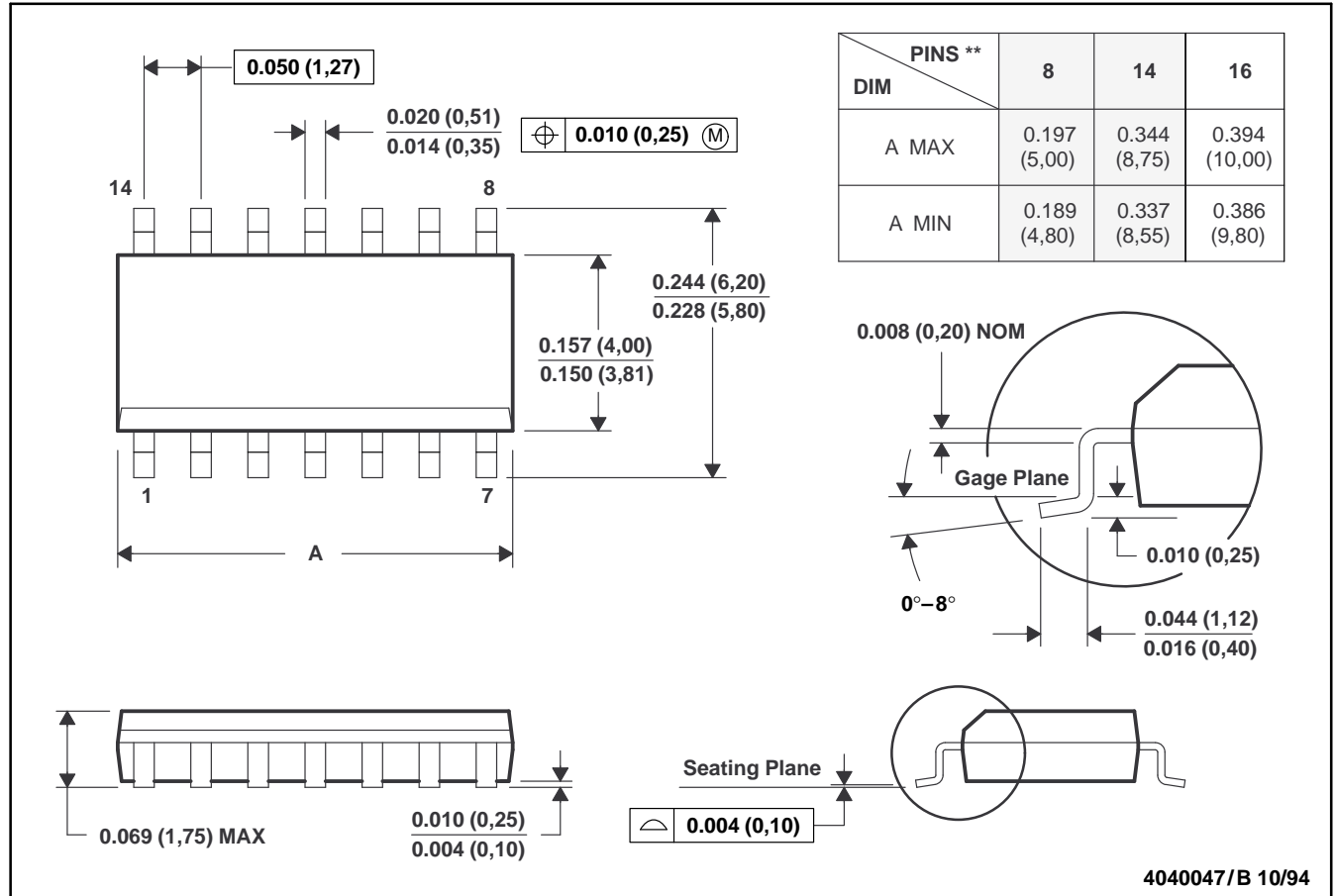
14 PIN SHOWN

PINS ** DIM	8	14	16
A MAX	0.197 (5,00)	0.344 (8,75)	0.394 (10,00)
A MIN	0.189 (4,80)	0.337 (8,55)	0.386 (9,80)

4040047/B 10/94

NOTES: A. All linear dimensions are in inches (millimeters).
        B. This drawing is subject to change without notice.
        C. Body dimensions do not include mold flash or protrusion not to exceed 0.006 (0,15).
        D. Four center pins are connected to die mount pad
        E. Falls within JEDEC MS-012

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MECHANICAL DATA

N (R-PDIP-T)**　　　　　　　　　　　　　　　　　　　　　　　**PLASTIC DUAL-IN-LINE PACKAGE**

16 PIN SHOWN
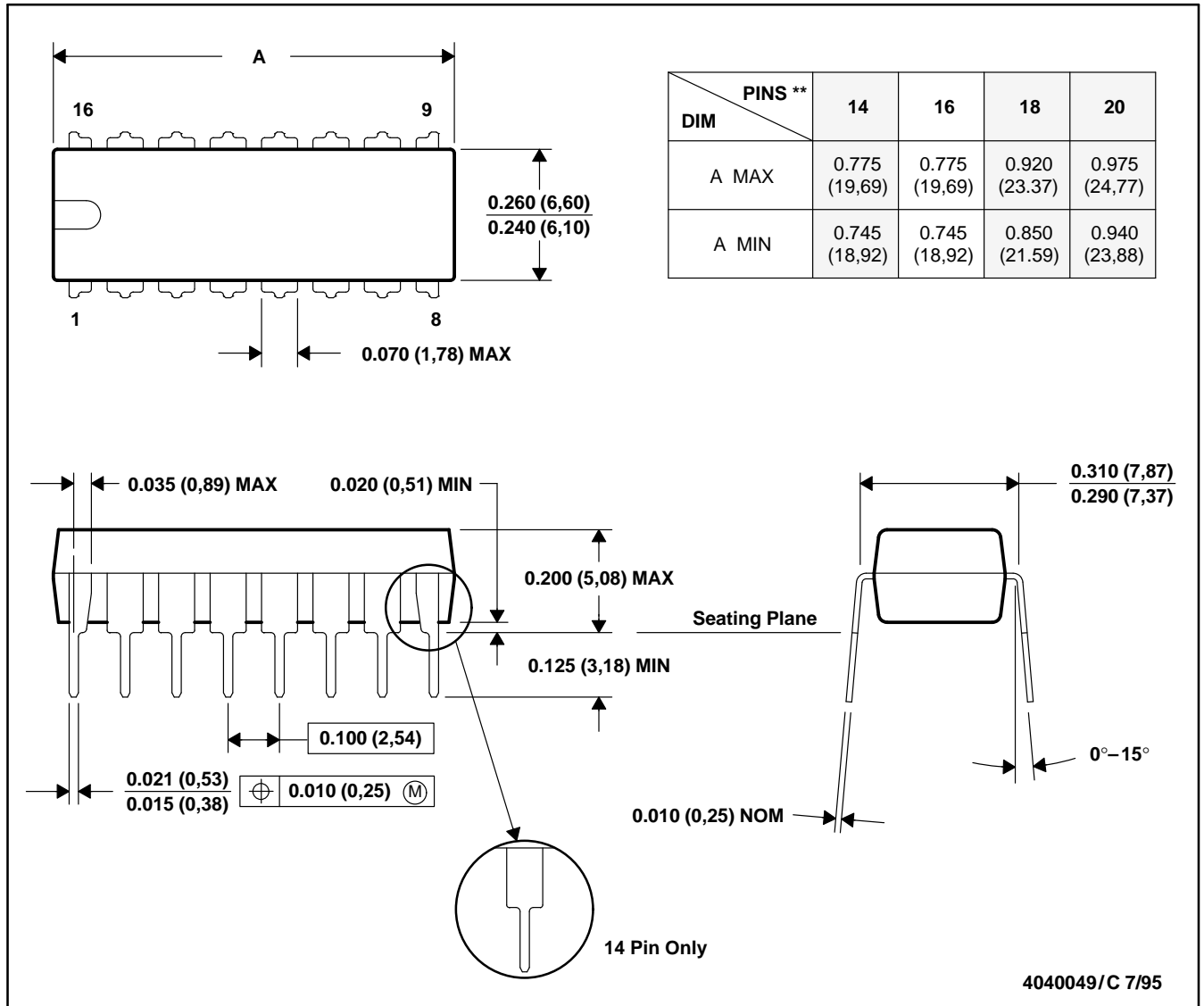
PINS ** DIM	14	16	18	20
A　MAX	0.775 (19,69)	0.775 (19,69)	0.920 (23,37)	0.975 (24,77)
A　MIN	0.745 (18,92)	0.745 (18,92)	0.850 (21,59)	0.940 (23,88)

4040049 / C 7/95

NOTES: A. All linear dimensions are in inches (millimeters).
B. This drawing is subject to change without notice.
C. Falls within JEDEC MS-001 (20-pin package is shorter than MS-001)

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MECHANICAL DATA

PW (R-PDSO-G**)　　　　　　　　　　　　　　　　　　　　　　　**PLASTIC SMALL-OUTLINE PACKAGE**

14 PIN SHOWN

PINS ** DIM	8	14	16	20	24	28
A MAX	3,30	5,30	5,30	6,80	8,10	10,00
A MIN	2,90	4,90	4,90	6,40	7,70	9,60
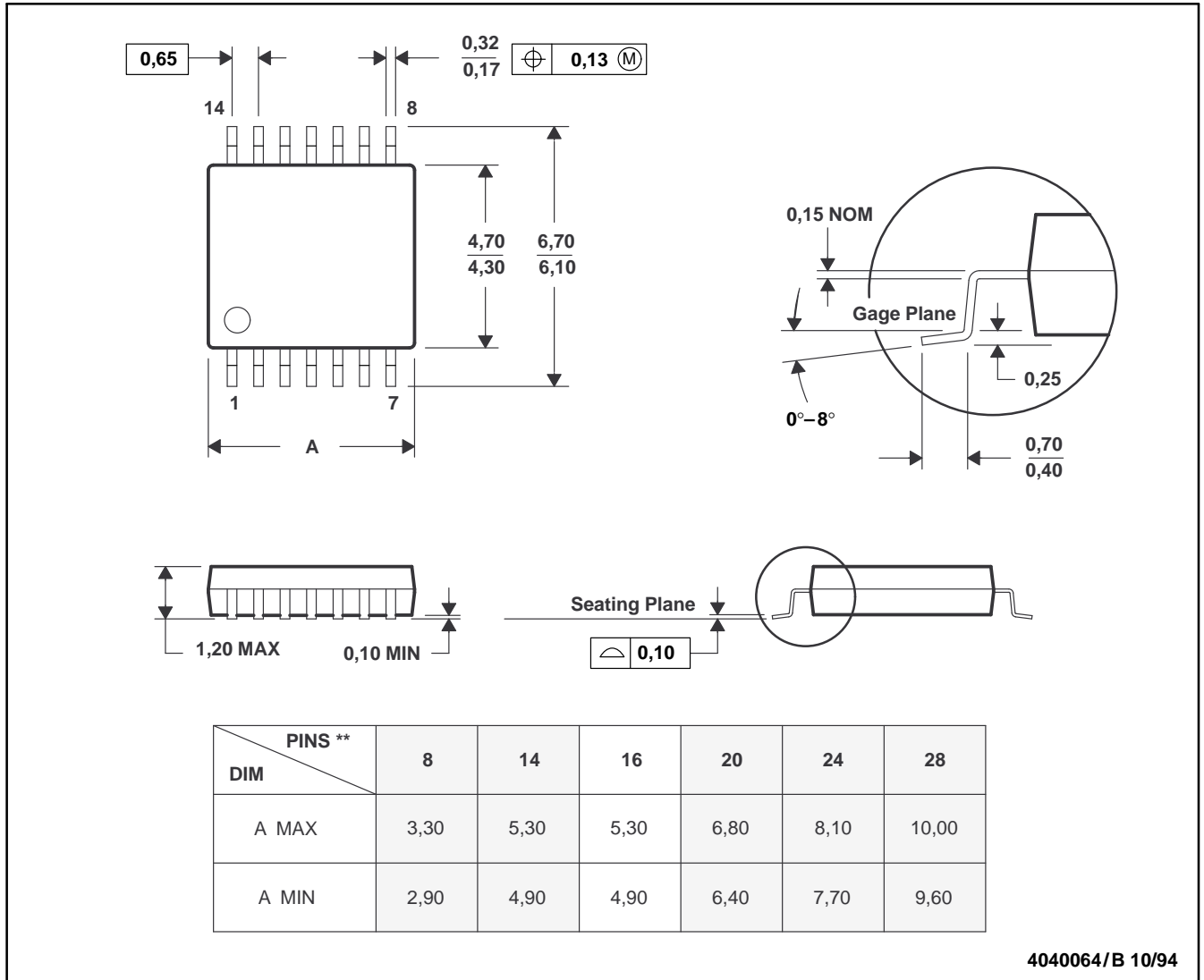
4040064 / B 10/94

NOTES:　A.　All linear dimensions are in millimeters.
　　　　　B.　This drawing is subject to change without notice.
　　　　　C.　Body dimensions do not include mold flash or protrusion not to exceed 0,15.

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